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CONCURRENT SPATIAL MAPPING OF THE VISCOELASTIC BEHAVIOR OF HETEROGENEOUS SOFT MATERIALS VIA A

POLYMER-BASED MICROFLUIDIC DEVICE

by

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> A Dissertation Submitted to the Faculty of Old Dominion University in Partial Fulfillment of the Requirements for the Degree of

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ABSTRACT

CONCURRENT SPATIAL MAPPING OF THE VISCOELASTIC BEHAVIOR OF HETEROGENEOUS SOFT MATERIALS VIA A POLYMER-BASED MICROFLUIDIC DEVICE

Wenting Gu Old Dominion University, 2016 Director: Dr. Julie Zhili Hao

This dissertation presents a novel experimental technique, namely concurrent spatial mapping (CSM), for measuring the viscoelastic behavior of heterogeneous soft materials via a polymer-based microfluidic device. Comprised of a compliant polymer microstructure and an array of electrolyte-enabled distributed resistive transducers, the microfluidic device detects both static and dynamic distributed loads. Distributed loads deform the polymer microstructure and are recorded as resistance changes at the locations of the transducers.

The CSM technique identifies the elastic modulus of soft materials by applying a precisely controlled indentation depth using a rigid probe to a sample placed on the device. The spatially-varying elastic modulus of the sample translates to a non-uniform load, causing a non-uniform deformation of the microstructure and variations in the recorded resistance changes. The CSM technique measures the loss modulus of soft materials through a dynamic measurement by applying varying sinusoidal loads to a sample placed on the device. The spatially-varying loss modulus of the sample causes the microstructure to respond with corresponding time delay. Consequently, the phase shift between the sinusoidal load and deflection of the sample along its length are captured by the distributed transducers.

As the first step of the experimental protocol, control experiments are implemented on the device to determine its static performance and system-level dynamic parameters. Next, the CSM technique is applied to both homogeneous and heterogeneous synthetic soft materials to measure their elastic moduli by applying a precisely controlled indentation depth through a probe, and the recorded load and device deflection are the output. The data are processed to obtain the overall load and the deflection of the sample at each transducer location and are further used to extract the elastic modulus distribution of the sample. The CSM technique is then applied to measure the loss modulus of soft materials. The measurable sinusoidal loads are the input, and the sinusoidal deflections of the device are the output. By applying the Fast Fourier Transform (FFT) algorithm and the nonlinear regression method, the data are processed to obtain the phase shift between the applied load and the device response along its microchannel length as well as the system-level parameters, namely stiffness (K), damping coefficient (D), and mass (M). In conjunction with the system-level parameters of the system with the device, obtained from the control experiment, the stiffness and the damping coefficient of a sample are calculated, and the sample's loss modulus distribution is estimated accordingly. This CSM technique successfully measures the spatiallyvarying elastic modulus and loss modulus of soft materials. As compared with the nanoindentation-based technique, the CSM technique demonstrates its efficiency in spatially mapping the viscoelastic behavior of a sample without excluding interactions among neighboring compositions in a sample.

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This dissertation is dedicated to my parents and grandmother.

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TABLE OF CONTENTS

		Page
LIST O	F TABLES	X
LIST O	F FIGURES	xii
СНАРТ	ER 1 INTRODUCTION	1
1.1	Methods for the Measurement of Viscoelastic Properties of Soft Materials	2
1.2	Micro/Nano Technology Utilized in the Measurement of the Viscoelastic Properties of Soft Materials	12
1.3	Mathematical Models for Extracting the Viscoelastic Properties of Materials using Micro/Nano Technology	19
1.4	Motivation	21
1.5	Objectives	22
1.6	Dissertation Layout	23
CHAPT LOAD	ER 2 A POLYMER-BASED MICROFLUIDIC DEVICE FOR DISTRIBUTED- DETECTION	24
2.1	Working Principle	24
2.2	Fabrication Process	37
2.3	Performance Characterization	42
2.4	Technical Issues Encountered	46
2.5	Conclusions	52
CHAPT SOFT N	ER 3 CONCURRENT SPATIAL MAPPING OF THE ELASTIC MODULUS OF MATERIALS: THEORY	54
3.1	Contact Mechanics	54
3.2	Rationale	60
3.3	Finite Element Analysis	62
3.4	Discussion and Conclusions	68
CHAPT SOFT M	ER 4 CONCURRENT SPATIAL MAPPING OF THE ELASTIC MODULUS OF MATERIALS: IMPLEMENTATION	70
4.1	Materials and Methods	70
4.2	Data Analysis	76
4.3	Results	77

	Page
4.4 Discussion and Conclusions	87
CHAPTER 5 CONCURRENT SPATIAL MAPPING OF THE LOSS MODULUS OF SOFT MATERIALS: THEORY	88
5.1 Rationale	88
5.2 Relation between the Loss Modulus and the Damping Coefficient	94
CHAPTER 6 CONCURRENT SPATIAL MAPPING OF THE LOSS MODULUS OF SOFT MATERIALS: IMPLEMENTATION	97
6.1 Materials and Methods	97
6.2 Data Analysis and Results	103
6.3 Discussion and Conclusions	115
CHAPTER 7 CONCLUSIONS AND FUTURE WORK	119
7.1 Discussion	119
7.2 Conclusions	119
7.3 Future Work	121
REFERENCES	124
APPENDICES	
A. CALIBRATION OF WENGLOR CP08MHT80 DISTANCE SENSOR	131
B. HEAT EQUILIBRIUM TEST FOR THE MICROFLUIDIC DEVICE	134
C. CALIBRATION OF ATI NANO17 LOAD CELL WITH WENGLOR®	135
D. DERIVATION OF THE VOLTAGE OUTPUT EQUATION OF THE	133
MICROFLUIDIC DEVICE	138
E. MEASUREMENT OF THE SPATIALLY-VATYING PHASE SHIFTS	1 40
WITH A DEVICE WITH A LAKUE TKANSDUCEK SPACING	140

F.	LABVIEW PROGRAMS FOR INSTRUMENT CONTROL AND DATA	
	ACQUISITION	152
G.	MATLAB CODES FOR DATA PROCESSING.	155
	VITA	159

LIST OF TABLES

Table Pag	e
2.1 Key design parameters of the microfluidic force sensor	5
2.2 Material properties used in the simulation	1
2.3 Resistance and voltage output of a device with varying microchannel height at the locations of the transducers	9
4.1 Thickness of the prepared samples	3
4.2 Coefficients that relate the indentation depth and the change of voltage output from the device	8
4.3 Elastic modulus obtained from 1:30 homogeneous PDMS sample	9
4.4 Elastic modulus obtained from the 1:30 and 1:40 heterogeneous PDMS samples through CSM	1
4.5 Measured elastic modulus of PDMS with 1:30 and 1:40 mixing ratios in the literature 8	1
4.6 Elastic modulus obtained from the homogeneous agar samples with different concentrations through the CSM technique	4
4.7 Elastic modulus of the agar samples with different concentrations in the literature	4
6.1 Dimensions of the testing samples	0
6.2 System-level parameters of the dynamic system with the microfluidic device and the setup, based on the linear curve fitting	7
6.3 System-level properties of the dynamic system	7
6.4 Measurement results of the dynamic system with a piece of testing material 10	9
6.5 Decoupled results of the testing materials	0
6.6 Measurement results of the samples after normalization with comparison to the information provided by Smooth-on Inc	1
6.7 Phase shifts of the two 1:30 PDMS heterogeneous samples at the frequency of 18 Hz and their normalized values	4
6.8 Estimation of the loss modulus distribution of the two 1:30 PDMS heterogeneous samples at the frequency of 18 Hz	4
6.9 System level parameters of the dynamic system based on the linear curve fitting when a different setup is used	6
A.1 Averaged voltage values as the micromanipulator staying still at each motion cycle	1
A.2 Ratio between voltage change and displacement change between two neighboring cycles	3
B.1 Voltage output from the PCB	4

Table	Page
C.1 Comparison of force amplitudes between experimental and theoretical values under different vibration amplitude.	137
E.1 Dimension of the homogeneous samples.	142
E.2 Dimension of the heterogeneous samples and the heterogeneity position	142
E.3 The maximum standard deviation of the measurement for different homogeneous samples below 80 Hz.	145

xii

LIST OF FIGURES

Figure Pag	ge
1.1 Schematic of the sample testing setup [34]	. 3
1.2 Schematic illustration of the AFM scanning process	. 6
1.3 Schematic of the indentation process by AFM probe [25].	. 7
1.4 Schematic of (a) a parabolic tip and (b) a conical tip	. 8
1.5 Force-indentation depth curves taken from PG-depleted and untreated cartilages using two different tips for the AFM-based nanoindentation [27]	10
1.6 (a) Relation between force, indentation depth and time, when an untreated cartilage disk is under dynamic indentation with a spherical probe tip in Phosphate Buffered Saline (PBS), and (b) the obtained storage and loss modulus [27]	11
1.7 Structure of a V-shaped thermal actuator array [43].	13
1.8 Schematic of a tactile force sensor that can detect force location [44]	13
1.9 Schematic and photo of a piezoresistive-based tactile sensor aiming for minimally invasive surgery [45]	14
1.10 Schematic of a resonator device for in vivo measurement of regional tissue viscoelasticity [46]	15
1.11 Material elastic property measurement of a specimen using a microfluidic device [47] 1	16
1.12 (a) Quantifying the elastic property of a testing sample and (b) quantifying the transient strain response of a 2% gellan gum and an S. epidermidis biofilm [47]	16
1.13 Schematic of a MEMS resonant sensor to characterize the viscoelastic properties of hydrogels [48]	17
1.14 Schematic of a MEMS microgripper for measuring the viscoelastic properties of soft hydrogel microcapsules [49]	18
1.15 Schematic of a spherical cell being grabbed by flat clippers, where the Hertz model is also applied [49]	20
2.1 Schematic view of the microfluidic force sensor	24
2.2 One sensing segment of the microfluidic device for distributed force sensing (a) side view of the segment (b) deformation of the polymer structure (c) geometry change of the electrolyte and its corresponding resistance change.	26
2.3 Photos of (a) a cylindrical probe [60] and (b) a rectangular probe	28
2.4 The circuit configuration and signal flow of the microfluidic device for obtaining a DC voltage output	29

Page

2.5 Simulation result of the displacement field at the y direction for (a) device #1 with a thinner PDMS structure and (b) device #2 with a thicker PDMS structure (A and B represent the displacement applied by the probe, and A' and B' represent the microchannel top deformation.).	33
2.6 Simulation result of the deformation of the microchannel top under different indentation depth applied by a rigid probe for both devices	34
2.7 Simulation result of the relation between the reaction force and the indentation depth under a rigid probe for both devices.	35
2.8 Simulation result of the relation between the resistance and indentation depth of one sensing segment for both devices.	36
2.9 Simulation result of the relation between the voltage output and indentation depth of one sensing segment for both devices	36
2.10 Deformation of the microchannel top of one sensing segment of the large microfluidic device when the indentation depth is 100 μm	37
2.11 Mask design for (a) the microchannel and (b) the patterned electrodes.	38
2.12 Schematics of (a) an SU-8 mold for making the microchannel and two reservoirs on a Pyrex slide, and (b) electrodes patterning with S1800 photoresist on another Pyrex slide.	39
2.13 Au/Cr electrodes on the Pyrex slide by sputtering and metal deposition.	40
2.14 SU-8 mold of the microchannel being covered by the 10:1 PDMS mixture cured at room temperature.	41
2.15 Fabricated microfluidic device with electrolyte in the microchannel	42
2.16 Relation between (a) force and indentation depth and (b) voltage output from five transducers and indentation depth of device #1 (large device)	43
2.17 Relation between (a) force and indentation depth and (b) voltage output from five transducers and indentation depth of device #2 (small device).	44
2.18 (a) Schematic of the z-direction misalignment of a probe (front view) and (b) simulation result of microchannel top deformation of device #1 under the z-direction misalignment of the probe, when the applied displacement equals to 60 μm.	47
2.19 (a) Schematic of the in-plane misalignment of the probe (top view) and (b) the effect of the in-plane misalignment on the microchannel top deformation under 60 μm prescribed displacement.	48
2.20 Schematic of the misalignment about the x-axis (side view).	48
2.21 Schematic of a device with variations in microchannel height (front view)	49
2.22 Initial voltages of the device output measured within one week	51
2.23 Voltage output of the device at its vibration equilibrium point measured during one dynamic test.	51

3.1 Schematic of a flat sheet under the indentation of a cylinder	55
3.2 A finite element model for verifying the relation between the reaction force and the indentation depth exerted by a cylinder	56
3.3 Theoretical and FEA simulation results of the contact model of a cylinder and a flat sheet with infinite thickness	57
3.4 Simulation results of force-indentation depth relation with different sheet thicknesses	58
3.5 The relation between the correction factor α and the thickness of a flat sheet	59
3.6 Schematic of the side view of one piece of soft material under indentation applied by a cylindrical probe.	60
3.7 Rationale of the CSM technique for the elastic modulus measurement	61
3.8 Simulation of a piece of heterogeneous material under the indentation applied by a cylindrical probe on a microfluidic force sensor.	63
3.9 Simulated microchannel top deformation when a piece of homogeneous material is under the indentation applied by a cylinder probe on the microfluidic device (The dashed lines represent the locations of the transducers).	64
3.10 Simulated microchannel top deformation when a piece of heterogeneous material under compression by a cylinder probe on the microfluidic force sensor (The dashed lines represent the locations of the transducers).	65
3.11 Simulation of a piece of material sample with a hard sphere (r = 1.2 mm) embedded, indented by a cylindrical probe on the microfluidic device	65
3.12 Deformation of the microchannel top along the microchannel length, when the distance between the embedded sphere and the device surface increases (The dashed lines represent the locations of the transducers).	66
3.13 Deformation of the microchannel top along the microchannel length, when the radius of the embedded sphere increases (The dashed lines represent the locations of the transducers).	67
3.14 Deformation of the microchannel top along the microchannel length, when the Young's modulus of the embedded sphere increases (The dashed lines represent the locations of the transducers).	68
4.1 Experimental setup for conducting concurrent spatial mapping	71
4.2 A polymer-based microfluidic device with 30 mm microchannel length and a PDMS container to accommodate materials needing to be tested in an aqueous condition [80]	71
4.3 Prepared (a) 1:30 homogeneous PDMS (b) 1:30 heterogeneous PDMS and (c) 1:40 heterogeneous PDMS samples [80]	73
4.4 Displacement pattern of the probe for conducting device characterization	75
4.5 Flow chart of the data processing procedure of the CSM of the elastic modulus of soft materials.	76

Page

4.6	Force and indentation depth relation of the 1:30 homogeneous PDMS sample and the fitting curves.	79
4.7	Force and indentation depth relation of the 1:40 heterogeneous PDMS sample and the fitting curves.	80
4.8	Measured elastic modulus of the homogeneous (a) 1% agar, (b) 2% agar, and (c) 3% agar samples as a function of indentation depth	82
4.9	Measured elastic modulus of the heterogeneous (a) 1% agar, (b) 2% agar, and (c) 3% agar samples as a function of the indentation depth.	85
5.1	Illustration of phase shifts between stress and strain of (a) purely elastic material, (b) viscous material and (c) viscoelastic material under harmonic excitation.	89
5.2	Schematic of the rationale for measuring the heterogeneous viscous behavior of a piece of soft material.	90
5.3	(a) Schematic of the side view of a microfluidic device under external sinusoidal loading, and (b) a second-order mechanical system model of the polymer-based microfluidic device under a sinusoidal excitation.	91
5.4	(a) Schematic of the side view of a piece of soft material under the dynamic measurement, and (b) a second-order mechanical system model of a piece of soft material under a dynamic measurement.	93
6.1	Experimental setup	97
6.2	A microfluidic device used in the loss modulus measurement of soft materials through CSM.	98
6.3	Schematic of the experimental setup and signal flow for the loss modulus measurement of soft materials through CSM	99
6.4	Photos of the samples: (a) dragon skin 10, (b) dragon skin 20, (c) Ecoflex, (d) dragon skin FX-Pro, (e) the first 1:30 PDMS sample with heterogeneity inside, and (f) the second 1:30 PDMS sample with heterogeneity inside	101
6.5	Phase and the amplitude ratio obtained from the dynamic characterization of the system with the setup and the microfluidic device.	105
6.6	Dynamic stiffness of the system as a function of $\omega 2$	106
6.7	Dynamic damping of the system as a function of ω	106
6.8	Dynamic measurement results of (a) phase and frequency relation, and (b) compliance and frequency relation of the dynamic system with Dragon Skin 10	108
6.9	Calculated loss modulus of the testing materials	112
6.1	0 Phase shift between the force and the voltage output from the device with a 0.75 mm transducer spacing, when (a) heterogeneous 1:30 PDMS sample1, (b) heterogeneous 1:20 PDMS sample2 are tested	112
C 1	1:50 FDINIS sample2 are tested.	113
0.1	requency response of the free vibration of the snaker and probe.	11/

Page

7.1 Conceptual drawings of (a) a portable force sensor with distributed force sensing element [44] and (b) conventional forceps mounted with a soft tactile sensor array [101]		
 A.1 Voltage output from the disptance sensor. 132 C.1 Setup for conducting dynamic calibration of the load cell. 136 D.1 Configuration of the circuits. 138 E.1 Experimental setup for measuring the spatially-varying phase shifts. 140 E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output). 143 E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency. 144 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample. 146 E.6 Relation between phase shift and frequency of a 1% agar sample. 147 E.8 Relation between phase shift and frequency of a 3% agar sample. 147 E.9 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 148 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.11 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.12 Relation between phase shift and frequency of a 1:40 heterogeneous agar sample. 149 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 149 E.142 Relation between phase shift and frequency of a 1:40 heterogeneous agar sample. 150 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection. 153 F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude. 	7.1 Conceptual drawings of (a) a portable force sensor with distributed force sensing element [44] and (b) conventional forceps mounted with a soft tactile sensor array [101].	. 123
 C.1 Setup for conducting dynamic calibration of the load cell. 136 D.1 Configuration of the circuits. 138 E.1 Experimental setup for measuring the spatially-varying phase shifts. 140 E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output). 143 E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency. 144 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample. 146 E.6 Relation between phase shift and frequency of a 1% agar sample. 147 E.8 Relation between phase shift and frequency of a 2% agar sample. 147 E.9 Relation between phase shift and frequency of a 1:30 heterogeneous PDMS sample. 148 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.11 Relation between phase shift and frequency of a 1:40 heterogeneous agar sample. 149 E.12 Relation between phase shift and frequency of a 1% heterogeneous agar sample. 149 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection. 154 	A.1 Voltage output from the disptance sensor	. 132
 D.1 Configuration of the circuits. 138 E.1 Experimental setup for measuring the spatially-varying phase shifts. 140 E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output). 143 E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency. 144 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample. 146 E.6 Relation between phase shift and frequency of a 1:40 homogeneous PDMS sample. 146 E.7 Relation between phase shift and frequency of a 2% agar sample. 147 E.8 Relation between phase shift and frequency of a 3% agar sample. 148 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.11 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.12 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.12 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection. 153 F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude. 	C.1 Setup for conducting dynamic calibration of the load cell	. 136
 E.1 Experimental setup for measuring the spatially-varying phase shifts. 140 E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output). 143 E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency. 144 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample. 146 E.6 Relation between phase shift and frequency of a 1% agar sample. 147 E.8 Relation between phase shift and frequency of a 3% agar sample. 147 E.9 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 148 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.11 Relation between phase shift and frequency of a 1% heterogeneous agar sample. 149 E.12 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 149 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection. 154 	D.1 Configuration of the circuits.	. 138
 E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output). 143 E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency. 144 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample. 146 E.6 Relation between phase shift and frequency of a 1% agar sample. 146 E.7 Relation between phase shift and frequency of a 2% agar sample. 147 E.8 Relation between phase shift and frequency of a 1:30 heterogeneous PDMS sample. 147 E.9 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 148 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.11 Relation between phase shift and frequency of a 2% heterogeneous agar sample. 150 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection. 153 F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude. 	E.1 Experimental setup for measuring the spatially-varying phase shifts	. 140
 E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency. 144 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample. 145 E.5 Relation between phase shift and frequency of a 1:40 homogeneous PDMS sample. 146 E.6 Relation between phase shift and frequency of a 1% agar sample. 147 E.8 Relation between phase shift and frequency of a 3% agar sample. 147 E.9 Relation between phase shift and frequency of a 1:30 heterogeneous PDMS sample. 148 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample. 149 E.11 Relation between phase shift and frequency of a 2% heterogeneous agar sample. 149 E.12 Relation between phase shift and frequency of a 3% heterogeneous agar sample. 150 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection. 153 F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude. 	E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output).	. 143
 E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample	E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency.	. 144
 E.5 Relation between phase shift and frequency of a 1:40 homogeneous PDMS sample	E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample	. 145
 E.6 Relation between phase shift and frequency of a 1% agar sample	E.5 Relation between phase shift and frequency of a 1:40 homogeneous PDMS sample	. 146
 E.7 Relation between phase shift and frequency of a 2% agar sample	E.6 Relation between phase shift and frequency of a 1% agar sample	. 146
 E.8 Relation between phase shift and frequency of a 3% agar sample	E.7 Relation between phase shift and frequency of a 2% agar sample	. 147
 E.9 Relation between phase shift and frequency of a 1:30 heterogeneous PDMS sample	E.8 Relation between phase shift and frequency of a 3% agar sample	. 147
 E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample	E.9 Relation between phase shift and frequency of a 1:30 heterogeneous PDMS sample	. 148
 E.11 Relation between phase shift and frequency of a 1% heterogeneous agar sample	E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample	. 149
 E.12 Relation between phase shift and frequency of a 2% heterogeneous agar sample	E.11 Relation between phase shift and frequency of a 1% heterogeneous agar sample	. 149
 E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample	E.12 Relation between phase shift and frequency of a 2% heterogeneous agar sample	. 150
 F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection	E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample	. 150
F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude	F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection	. 153
	F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude	. 154

CHAPTER 1

INTRODUCTION

The study of the viscoelastic properties of soft materials can be of great impact in the engineering and biological fields [1-10]. For engineering applications, soft materials are widely employed to build mechanical structures for lab-on-chip microfluidic devices, tactile sensors, and wearable devices [1, 3, 4]. The ability to determine the viscoelastic properties of structural material is essential to the accuracy and stability of the sensor performance [4]. The hysteresis existing in most of the flexible sensors and the energy dissipation during the sensor operation are substantially associated with the viscoelastic properties [11]. In biological fields, the viscoelastic properties of biological materials are measured to determine the pathophysiology of diseases and the corresponding treatment [12-14]. Furthermore, studies have shown that human morbidity and mortality are profoundly related to the degradation and mechanical failure of biological tissues, indicated by the measured viscoelastic properties [5, 6]. Internal disease or injuries, such as breast cancer, atherosclerosis, fibrosis, and glaucoma can be indicated by changes in the macroscopic stiffness of tissues [15]. Even on the cellular level, cell morphology, proliferation, differentiation and motility would also be influenced by local compliance [16, 17]. The viscoelastic property reflects the tissue's ability to withstand physiological loading, checks for proper functionality and the integrity of the tissue, and serves as an important indication of tissue health [18, 19]. Conventional experimental techniques usually measure the overall elastic modulus or loss modulus at the macroscopic level [6, 20-23], which cannot capture the heterogeneity of the material, or they sequentially map the mechanical property at the microscopic level, which is time-consuming [24-28]. As such, the development of a new experimental technique to measure the spatially-varying elastic modulus and loss modulus in an efficient way, can be of great significance in assisting

material selection in the engineering field and tissue health diagnosis in the biological field.

1.1 Methods for the Measurement of Viscoelastic Properties of Soft Materials

Various experimental techniques have been developed to measure the viscoelastic properties of soft materials, at both the macroscopic and microscopic levels [15, 21, 24-32].

1.1.1 Methods for the viscoelastic property measurement at the macroscopic level

The Pulse Wave Velocity (PWV) method is an approach commonly used for measuring arterial stiffness and elastic modulus. The velocity of a pulse wave traveling in the vascular wall is measured and used to estimate the elastic modulus of the vascular wall [21]. Scanning Acoustic Microscopy (SAM) is a method that maps the elastic modulus of isolated cells and whole tissues, but it is limited to hard biological tissues, such as dentin and bone [6]. Magnetic Resonance Elastography (MRE) is another approach to characterize the mechanical property of tissues [20, 22]. Similarly, the Shearwave Dispersion Ultrasound Vibrometry (SDUV) determines the elastic modulus and loss modulus of tissues, by detecting the local deformation of tissues under a harmonic mechanical excitation of shear waves with frequencies up to 1000 Hz [23].

One major category of techniques for measuring the viscoelastic property of soft materials is through mechanical approaches. Krouskop *et al.* [29] investigated the viscoelastic behavior of breast and prostate tissue samples. A hydraulic testing machine was used to apply a uniaxial compressive sinusoidal load to the sample, and a load cell was assembled to measure the applied force. This study proved that breast fat tissue has a constant modulus regardless of the strain level, while the modulus of other tissues is dependent on the strain level. At a high strain level, carcinomas from the breast were found to be stiffer than the glandular and fibrous tissues. Similarly, cancerous prostate tissue exhibited the same pattern. The analytical model used in their research is based on a uniform load acting on a semi-infinite elastic solid. Kim *et al.* [33] measured

the compressive stiffness and radial streaming potential of a cartilage disk under dynamic compression loading in an unconfined compression chamber.

Optical methods could be incorporated with mechanical approaches to help determine the strain field within a sample [34, 35]. Buckley *et al.* [34] used high-speed confocal imaging techniques to accurately resolve the strain field of articular cartilage samples in both space and time domains. Along with the force sensor embedded in the shear plates, the high-resolution tissue strain mapping and force measurement are synchronized. Figure 1.1 shows the schematic of the test setup. The sample was sheared by two parallel plates, while it was imaged from below with a confocal microscopy. Through this setup, the moving plate exerted a sinusoidal shear force to the sample at frequencies of f = 0.01, 0.1 and 1 Hz, with a pre-applied 20% compressive strain perpendicular to the moving direction.



Figure 1.1 Schematic of the sample testing setup [34].

In their test, cartilage samples were cut into hemicylinder shapes and placed between two sand blasted glass-shearing plates. One of the plates was fixed and attached to a leaf spring, while the other could be actuated. The deep zone of the articular cartilage was glued to the fixed plate, and the surface was attached to the moving one. A photo-bleached sample was imaged by a Zeiss LSM 5 confocal microscopy with 20X magnification. Unstained samples were imaged with a Zeiss LSM 710 microscope along with a 488-nm laser for illumination. It was demonstrated that viscoelastic properties of articular cartilage vary drastically with depth, showing a location dependent mechanical property.

1.1.2 Instrument indentation methods for the measurement of viscoelastic behavior of materials at the microscopic level

The aforementioned methods all measure the viscoelastic properties at a macroscopic level. In a macroscopic testing, strain fields usually span the entire sample. Another mechanical approach that is commonly used for sample viscoelastic properties measurement is based on the forcedeformation response through indentation at the microscopic level [24, 26-28, 36]. Different from the macroscopic testing, indentation induced deformation only concentrates at the point where the indenter contacts the sample and gradually diminishes with the increasing distance from the indenter.

For the indentation type technique, a variety of instruments have been used for material indentation, ranging from atomic force microscope (AFM), and nanoindenters to larger industrial indenters. An AFM can apply loads as small as piconewton, nanoindenters can resolve nanonewton loads, while larger industrial indenters have a more flexible load range, from micronewtons to meganewtons. Depending on the material type and the specific research objective, different indentation methods are selected accordingly. Under an elastic deformation assumption, with the purpose of extracting the elastic modulus, it is crucial to fit the relation between the measured indentation depth and the indentation load, which is geometry specific [15, 30]. Due to its consistent and versatile procedure, instrumented nanoindentation has been used to quasi-statically

characterize the mechanical properties of a wide range of biological tissues, such as cartilage, enamel, vascular tissues and sclera [31]. In recent decades, dynamic indentation tests have been developed to better characterize the material properties, especially for measuring the loss modulus of materials. McLeod *et al.* [37] obtained the depth and directional dependence of the microscale biomechanical properties of porcine cartilage *in situ* via an AFM. Franke *et al.* [26] carried out a dynamic nanoindentation of articular porcine cartilage *in vitro* and compared the results at different load amplitudes over the same frequency range. With the development of these techniques, abnormal tissues are better studied. For example, musculoskeletal diseases on weight-bearing tissues are better understood based on the elastic modulus obtained with the help of nanoscale testing techniques [38].

The Atomic Force Microscope (AFM) was invented in 1986 by Binnig *et al.* [39]. Instrumental improvements and novel applications of AFM have developed rapidly over the last two decades. AFM has gradually become one of the most useful tools for studying local surface interactions as a Scanning Probe Microscope (SPM).

Figure 1.2 schematically illustrates the working principle of an atomic force microscope. The basic structural component for conducting the indentation via an AFM is a flexible cantilever with a microfabricated sharp tip. The tip will interact with sample surfaces during the measurement, and the other end of the cantilever is fixed to a holder. During the test, a laser beam is emitted from a laser diode onto the back surface of the cantilever, right above the tip. The laser beam is then reflected to a position-sensitive photodiode where the laser spot on the photodiode moves proportionally to the cantilever deflection. Assuming a constant stiffness of the cantilever, the interaction force between the tip and the sample can be obtained by multiplying the stiffness with the deflection. The sample is mounted on a piezoelectric scanner in the illustration above and can move along the x, y and z directions [40].



Figure 1.2 Schematic illustration of the AFM scanning process.

Figure 1.3 illustrates the deformation of a sample and the cantilever during an indentation test. As illustrated in Figure 1.3, the sphere represents the tip of an AFM's cantilever. From Figure 1.3, it can be seen that the entire translation of the beam Δz equals the summation of the cantilever deflection Δd and the indentation depth δ [25]. Δz can be precisely controlled by the piezoelectric scanner, while Δd is recorded by the photodiode. As such, the indentation depth on a sample is easily to be obtained by Equation 1.1:

$$\delta = \Delta z - \Delta d \tag{1.1}$$

where Δd can be further used to calculate the interaction force through $F = k \cdot \Delta d$, where k is the

stiffness of the cantilever, and is a constant value related to the material and geometry of the cantilever.



Figure 1.3 Schematic of the indentation process by AFM probe [25].

The Hertz model is most commonly used to relate the elastic modulus of the measured sample with the force-indentation curve obtained by an AFM static indentation measurement. However, many specific assumptions are required for the Hertz model to be applied. The tested sample is assumed to be an isotropic and linear elastic solid. The indentation is negligible in comparison to the sample thickness, which is 5-10% of the sample thickness maximum. Additionally, the indenter is assumed to be not deformable, and there are no additional interactions, such as adhesion between the indenter and the sample.

The selection of the cantilever for conducting the indentation depends on the stiffness of the test sample, which should be around the same range of the sample stiffness. For example, for delicate cells, the cantilever is chosen to be with a spring constant of around 10-30 mN/m [41].

The tip on the cantilever usually has a range of possible geometries. Based on these varying geometries, Hertz models with different mathematical expressions are applied.

For tips with a parabolic geometry, as shown in Figure 1.4 (a), the relation between force F and the indentation depth δ can be expressed as

$$F = \frac{4\sqrt{R_c}}{3} \frac{E}{1 - v^2} \delta^{3/2}, \qquad (1.2)$$

where R_c is the radius of the tip, E is the elastic modulus of the testing material and v is the Poisson's ratio.



Figure 1.4 Schematic of (a) a parabolic tip and (b) a conical tip.

Modifications were made based on the Hertz model to account for situations that do not actually meet the Hertz assumptions. Dimitriadis *et al.* [25] derived a theoretical model based on the Hertz model to better describe the relation between the force and indentation depth on thin samples, when a parabolic tip is applied.

$$F = \frac{4E\sqrt{R\delta^{3/2}}}{3(1-v^2)} \left(1 - \frac{2\alpha}{\pi} \chi + \frac{4\alpha^2}{\pi^2} \chi^2 - \frac{8}{\pi^3} \left(\alpha^3 + \frac{4\pi^2}{15} \beta \right) \chi^3 + \frac{16\alpha}{\pi^4} \left(\alpha^3 + \frac{3\pi^2}{5} \beta \right) \chi^4 \right)$$
(1.3)

where $\chi = \frac{\sqrt{R\delta}}{h}$, and *h* is the sample thickness.

Different coefficients α and β are used to account for conditions where the sample is adherent and non-adherent. For adherent samples:

$$\alpha = -0.347 \frac{3 - 2\nu}{1 - \nu} , \qquad (1.4)$$

$$\beta = 0.056 \frac{5 - 2\nu}{1 - \nu} ,$$

and for non-adherent samples:

$$\alpha = -\frac{1.2876 - 1.4678\nu + 1.3442\nu^2}{1 - \nu}$$

$$\beta = \frac{0.6387 - 1.0277\nu + 1.5164\nu^2}{1 - \nu}$$
(1.5)

For conical tips, as shown in Figure 1.4 (b), the force can be derived to be:

$$F = \frac{E}{1 - v^2} \frac{2 \tan \alpha}{\pi} \delta^2 \qquad (1.6)$$

where α is the semi-opening angle of the cone.

Similarly, improvements were made to the model to make it applicable to more complicated situations. Gavara *et al.* [42] derived a model to account for samples with a small thickness that are indented by a conical tip.

$$F = \frac{8E\tan\alpha}{3\pi}\delta^2 \left[1 + \beta \frac{2\tan\alpha}{\pi^2 h}\delta + 16\beta^2 \frac{\tan^2\alpha}{h^2}\delta^2\right]$$
(1.7)

where β is 1.7795 or 0.388 for adherent or non-adherent cases, respectively.

Other than the geometries mentioned above, tips with other geometries are also commonly used for the cantilever tip of an AFM. The geometry could be a pyramid, hyperboloid, blunt cone, blunt pyramid, truncated cone and truncated pyramid, etc. The choice of the indenter shape depends on the mechanical property of the material under measurement. For soft biological samples with a low elasticitic modulus, spherical probes are recommended, since the stress caused by the indenter will not be able to damage delicate samples. However, this type of tip may not be able to obtain high-resolution results. In this case, sharp tips, such as pyramidal silicon nitride tips,

are

employed because of their smaller contact dimension. However, sample penetration that will lead to inaccurate calculation should be avoided [41].

Han *et al.* [27] carried out an AFM-based classical nanoindentation in Phosphate Buffered Saline (PBS), using a maximum load of approximately 70 nN in the force mode. The closed looped control system in the z-direction enabled the precise control of both the indentation force and depth. Constant displacement rates of $0.1-10 \,\mu$ m/s along the z-axis were used. For each indentation curve, indentation force-depth relation data were obtained, as can be seen in Figure 1.5. Three cartilage disk samples, proteoglycan (PG)-depleted and untreated, were tested with both spherical and pyramidal probe tips. All the indentations were performed at different locations in relatively flat regions. At each location, the indentation was repeated, and between each repeat, the sample was rested until recovery.



Figure 1.5 Force-indentation depth curves taken from PG-depleted and untreated cartilages using two different tips for the AFM-based nanoindentation [27].

From the testing results of the classical indentation, it is noted that the indentation modulus E increases significantly with the increment of the displacement rate, regardless of the tip type.

For untreated cartilages, the indentation modulus increases from approximately 0.14 MPa to approximately 0.2 MPa; for PG-depleted cartilages, the increase is not as much, from around 0.5 MPa to 0.6 MPa. All the results fit the Hertz model well.

The same group also carried out a dynamic indentation test, as shown in Figure 1.6 (a). The sinusoidal displacement along the z-axis was implemented at the amplitudes of 4, 25, and 125 nm. Based on the Hertz model, the relation between the force amplitude, indentation amplitude, offset indentation depth, and the complex modulus were obtained, for the spherical and pyramidal tips, respectively. The storage and loss modulus were then obtained accordingly, as shown in Figure 1.6 (b).



Figure 1.6 (a) Relation between force, indentation depth and time, when an untreated cartilage disk is under dynamic indentation with a spherical probe tip in Phosphate Buffered Saline (PBS), and (b) the obtained storage and loss modulus [27].

1.2 Micro/Nano Technology Utilized in the Measurement of the Viscoelastic Properties of Soft Materials

Although the nanoindentation-based technique exhibits its advantages in measuring the viscoelastic behavior of a material on a submicrometer scale and is broadly applied in measuring the elastic modulus of biological materials, it also has drawbacks. For example, this technique only provides a local mechanical stimulation around 100 nm in depth, which fails to capture the indepth mechanical property. The stimulation is also limited by the size of the cantilever tip of an AFM. The interaction among neighboring compositions in a tissue specimen under physiological loading is also missing, especially at the tissue level study of biomechanics. Furthermore, the spatial mapping for a piece of material using AFM is time consuming and costly.

Other mechanical methods based on micro/nano technology have also been explored, due to their low cost and ability to integrate microsensors on a single chip. Zhang *et al.* [43] developed a V-shaped polymer electrothermal actuator array for measuring the mechanical compliance of a biological cell, as shown in Figure 1.7. The actuation mechanism has an electrothermal effect, as it enables the device to be operated in a cell medium. Dielectrophoresic quadrupole electrodes are used to trap a cell between a plunger and a force sensor. Then, the desired strain is applied to the cell by the V-shaped electrothermal actuator array while the force sensor measures the applied force. Meanwhile, the temperature during the experiment is monitored by a thermal sensor on the chip.



Figure 1.7 Structure of a V-shaped thermal actuator array [43].

Sokhanvar *et al.* [44] developed a tactile sensor that is able to measure force, force position, and the softness of a grasped object, as shown in Figure 1.8. A uniaxial polyvinylidene fluoride (PVDF) film is used as the transduction element. When soft materials are grasped, the bending of a PVDF beam generates a voltage signal, which is related to the deformation of the beam and is further related to the softness of the grasped object. As a drawback of piezoelectric materials, static loads are not able to be detected. Thus, a dynamic load is applied in the test and the peak-to-peak voltage value of the PVDF output is further analyzed.



Figure 1.8 Schematic of a tactile force sensor that can detect force location [44].

Similarly, Atieh [45] developed a piezoresistive-based tactile sensor, aiming for Minimally Invasive Surgery (MIS), which successfully measured the Young's modulus of silicone rubbers. The schematic of the tactile sensor and the photo of its prototype are shown in Figure 1.9.



Figure 1.9 Schematic and photo of a piezoresistive-based tactile sensor aiming for minimally invasive surgery [45].

There is sparse literature that highlights MEMS devices developed to detect the loss modulus of soft materials.

Dhar [46] *et al.* developed a resonator device for *in vivo* measurement of regional tissue viscoelastic properties, as can be seen in Figure 1.10. The main structure of this resonator comprises of one proof mass, one tip mass, and two springs between each. The tip mass will be in direct contact with the material that is going to be measured. The vibration is generated by a piezo actuator. The elastic modulus of the contact material is determined by the frequency shift from the original resonance, and the viscous component of the testing material at the contact region is

determined by the bandwidth of the frequency response.



Figure 1.10 Schematic of a resonator device for *in vivo* measurement of regional tissue viscoelasticity [46].

Researchers have also started to use microfluidic devices for the viscoelastic properties measurement of soft materials. Hohne *et al.* [47] measured bacterial biofilms, whose material property cannot be easily measured by conventional methods. As shown in Figure 1.11, the device is comprised of air chambers and fluidic channels with a thin PDMS layer between them. The testing sample is placed in the fluidic channel, while the pressure in the air chamber can be controlled to deform the PDMS layer, as well as the testing sample in the channel. A confocal microscope is utilized to monitor the deformation of the PDMS layer. Then, the deformation of the PDMS layer captured by a microscope is compared with finite element analysis (FEA) simulation results, as shown in Figure 1.12 (a). When the two curves match each other, the elastic modulus of the testing material is then determined. The transient response of the strain of the testing material is shown in Figure 1.12 (b), from which the viscous behavior is captured.



Figure 1.11 Material elastic property measurement of a specimen using a microfluidic device [47].



Figure 1.12 (a) Quantifying the elastic property of a testing sample and (b) quantifying the transient strain response of a 2% gellan gum and an S. epidermidis biofilm [47].

As shown in Figure 1.13, Corbin *et al.* [48] developed a MEMS resonant sensor to measure the viscoelastic properties of hydrogels over a range of concentrations. Gels with different concentrations were fabricated directly onto the devices using an electrohydrodynamic jet printing technique. The viscoelastic properties of the material is measured based on the technique of measuring frequency shift between a sensor without loading a hydrogel and a sensor loaded with a piece of hydrogel.



Figure 1.13 Schematic of a MEMS resonant sensor to characterize the viscoelastic properties of hydrogels [48].

Kim *et al.* [49] reported a MEMS microgripper which integrates two capacitive force sensors and one electrothermal microactuator on a single chip, as shown in Figure 1.14. The microgripper is capable of measuring the viscoelastic properties of soft hydrogel microcapsules at the size of around 15-25 μ m. The electrothermal microactuator generates gripping displacements at the arm tips, while the capacitive force sensors measure the force applied on the microcapsules and provide a feedback. Meanwhile, the material deformations are measured from phase contrast

images. The force-displacement and time-dependent displacements data are collected for the microcapsules. The viscoelastic properties are extracted based on a modified Hertz model accommodating large deformations.



Figure 1.14 Schematic of a MEMS microgripper for measuring the viscoelastic properties of soft hydrogel microcapsules [49].

With the help of a MEMS mirror, Lee [50] used an optical coherence tomography (OCT) method to capture the deformation of brain tissues of rats, under the press of a stainless steel bead with 1 mm in diameter. From there, the time-dependent deformation was recorded. A finite element model was also built in Abaqus, and was used to estimate the mechanical property by

comparing the simulation results with the deformation profile obtained from the optical measurement.

1.3 Mathematical Models for Extracting the Viscoelastic Properties of Materials using Micro/Nano Technology

The most commonly used mathematical model for extracting the elastic modulus from the measurement is the Hertz model. The Hertz model is widely used in cellular level material property measurement using AFM or nanoindentation technique, as mentioned previously [27, 41, 42]. In the above-mentioned examples, one sample is pressed by a spherical indenter tip. The same mathematical model is also applied to the scenario of one spherical cell being grabbed by flat clippers, as can be seen in Figure 1.15. In both cases, the measurement of the force-indentation relation is needed. From the relation, the elastic modulus of the testing material is calculated using Equation 1.8.

$$E = \frac{3F(1-\nu^2)}{4\sqrt{R\delta^3}} \tag{1.8}$$

where *F* is the measured force, δ is the measured indentation depth, and *R* is the radius of either the indenter tip or the spherical cell being grabbed.


Figure 1.15 Schematic of a spherical cell being grabbed by flat clippers, where the Hertz model is also applied [49].

Instead of using the classic Hertz model which only accounts for the force-displacement relation under small deformation, Kim [49] adopted the model developed by Tatara [51], which takes large deformation of soft material samples into consideration. As shown in Equations 1.9 and 1.10, higher order terms are added to relate force and deformation. As such, the elastic modulus of the material is achieved by curve fitting.

$$\delta = \frac{3(1-\nu^2)F}{4Ea} - \frac{f(a)F}{\pi E}$$
(1.9)

$$f(a) = \frac{2(1+\nu)R^2}{(a^2+4R^2)^{3/2}} + \frac{1-\nu^2}{(a^2+4R^2)^{1/2}}$$
(1.10)

where *a* is the contact radius, δ is the sample deformation, *R* is the radius of the cell, *v* is the Poisson's ratio, and *F* is the force applied on the spherical cell.

Peng [52] developed a theoretical model for extracting the elastic modulus of the testing materials, based on a MEMS tactile sensor developed by their group. Under the assumption that the deformation of a sample only occurs over a very small area, Equation 1.11 is developed.

$$E_{t} = \frac{2(1-\nu^{2})a}{L} \left(\frac{\frac{q_{h}}{q_{s}}-1}{\frac{1}{E_{s}}-\frac{1}{E_{h}}\frac{q_{h}}{q_{s}}}\right)$$
(1.11)

In Equation 1.11, E_h and E_s are known values that represent the values of Young's modulus of the hard and soft sensing nodes of the tactile sensor, q_h and q_s represent the force per unit area on the soft and hard sensing nodes, respectively. a is the radius of the loading zone, and Lrepresents the sample thickness.

The mathematical model that is commonly used to capture the viscous behavior of a material is the Prony series when a linear viscoelastic material is assumed [49, 53, 54], as expressed in Equation 1.12.

$$E_{relax}(t) = E_0 + E_1 \exp(-t/\tau_1) + E_2 \exp(-t/\tau_2)$$
(1.12)

where E_0 is the time-independent elastic modulus, E_1 , E_2 are the relaxation moduli, and τ_1 , τ_2 are the relaxation time constants.

1.4 Motivation

Mathematical models for extracting the viscoelastic properties of soft materials have been maturely developed for conventional material property measurement techniques, using instruments such as a nanoindenter or AFM. However, these methods are costly and used only within the nano scale or up to several microns [41, 55]. Sequentially mapping the viscoelastic properties of an area of the testing material can be time consuming.

At the macroscopic level, most mechanical approaches to measure the viscoelastic properties of soft materials only obtain one overall value [29, 33]. Portable, small-scaled devices are also developed for the viscoelastic properties measurement of soft materials, based on the micro/nano technology [43, 44, 47, 48]. Researchers have already paid great interests to polymer-based devices in the application of force detection, texture recognition or elastic modulus

measurement of soft materials including biological samples [47, 56-58], due to the flexibility of polymer materials and their biocompatibility. Only a few researchers utilized conductive liquid as a transduction mechanism. Gutierrez et al. [56] developed an impedance-based force sensor with a fluid-filled parylene microstructure. However, the nonlinear relations of the force and the device deflection, as well as the impedance change and the device deflection, caused the inconvenience in interpreting the data in real applications. Thus, no further quantitative application was discussed. Only a few of the works achieved a quantitative relation between the viscoelastic properties of the testing materials and the sensor output [47, 57], let alone quantifying the spatially-varying mechanical properties with one measurement. Ahmadi et al. [59] reported on a discretely loaded beam-type optical fiber tactile sensor which is able to detect the existence of a lump in a tissue specimen, when a tissue specimen is pressed against the sensor. To the best knowledge of the author, this work is the closest to the concept that to examine the spatially-varying elastic modulus of a soft material through one single device. However, no quantitative relation between the elastic property of the testing material and the device output is presented in the work. The miniaturization and mass production of the sensor is also a challenge, due to the size of the prototype (45 mm \times 8 $mm \times 8$ mm). On the other hand, other than measuring the spatially-varying viscoelastic properties by sequential measurement through the nanoindentation-based technique, no other groups have studied on measuring the spatially-varying viscoelastic properties through one device in one measurement.

1.5 Objectives

This work aims to develop a novel experimental technique to measure the spatially-varying viscoelastic properties of soft materials through a polymer-based microfluidic device developed for this study.

1.6 Dissertation Layout

Chapter Two presents the microfluidic device employed in the CSM in this work. The working principle and its performance characterization are presented. Technical issues associated with the device and the CSM technique are also discussed.

Chapter Three presents the theory of the CSM technique for acquiring the spatially-varying elastic modulus of soft materials.

Chapter Four presents the implementation of the CSM technique on the elastic modulus measurement of synthetic soft materials. The testing results are presented and compared with those in the literature. Heterogeneity within a material is also successfully detected.

Chapter Five presents the theory of the CSM technique for acquiring the spatially-varying viscous property of soft materials. The relation between the loss modulus and the measurable damping coefficient is also derived.

Chapter Six presents the implementation of the CSM technique for the viscous measurement of soft materials, including the spatially-varying phase shifts, the overall stiffness, the damping coefficient, and the frequency-dependent loss modulus.

Finally, Chapter Seven discusses the advantages and drawbacks of the CSM technique, and proposes future work that can be built upon this dissertation.

CHAPTER 2

A POLYMER-BASED MICROFLUIDIC DEVICE FOR DISTRIBUTED-LOAD DETECTION

This chapter presents a polymer-based microfluidic resistive sensor employed in this study for the measurement of the viscoelastic properties of soft materials. The sensor is mainly comprised of a rectangular polymer microstructure and an array of electrolyte-enabled distributed resistive transducers. Distributed loads deform the polymer microstructure and are recorded as resistance changes at the locations of the transducers. Since the detailed design, fabrication, and characterization process of the device have been presented in the previous work [60-62], a brief introduction of the device is given in this chapter for completeness. Then, the working principle, input-output relation, and technical issues associated with the microfluidic device and the experimental technique are elaborated in the rest of the chapter.

2.1 Working Principle

Figure 2.1 shows the configuration of the microfluidic device for distributed-load detection.



Figure 2.1 Schematic view of the microfluidic force sensor.

The main structure of the device is a rectangular polymer (Polydimethylsiloxane: PDMS) microstructure embedded with a microchannel. The microchannel is filled with an electrolyte, 1-ethyl-3-methylimidazolium dicyanamide (EMIDCA). Five electrode pairs are immersed in the electrolyte in the microchannel, functioning as five discrete resistive transducers together with the electrolyte. Two reservoirs, located at the two ends of the microchannel, allow the conductive liquid to be injected, stored, and to flow in and out during operation. The microstructure converts distributed normal loads to continuous deflections, which further convert to resistance changes at the transducer locations. The microchannel length, microchannel width, transducer spacing, and the polymer structure thickness are the key design parameters and that can be adjusted. The key parameters of the two devices employed in the CSM technique in this work are listed in Table 2.1.

Symbol	Device #1	Device #2	Notes
l_c	30 mm	12 mm	Length of the microchannel
Wc	1 mm	0.5 mm	Width of the microchannel
h_c	80 µm	80 µm	Height of the microchannel
<i>l_{PDMS}</i>	50 mm	25 mm	Length of the PDMS structure
WPDMS	20 mm	8 mm	Width of the PDMS structure
h_{PDMS}	1.2 mm	2 mm	Height of the PDMS structure
r_P	0.8 mm	0.8 mm	Radius of the probe for force exertion
l_P	11 mm	11 mm	Length of the probe for force exertion
$d_{e\!f\!f}$	1.5 mm	0.75 mm	Transducer spacing

Table 2.1 Key design parameters of the microfluidic force sensor.

To better illustrate how distributed normal loads are converted to the recordable resistance changes at the transducer locations, one sensing segment that includes one transducer and its corresponding mechanical structure is shown in Figure 2.2. Figure 2.2 (a) shows the cross section of the device, with the labels that represent its dimension listed in Table 2.1. Figure 2.2 (b) illustrates the deformation of one sensing segment. The electrolyte in that portion of the

microchannel functions as a variable resistor adjusted by the external load, as can be seen in Figure 2.2 (c). Under an external load, the resistance of the electrolyte portion changes due to the alteration of its geometry. PDMS is chosen as the structural material, due to its low elastic modulus and great fabrication simplicity. The ionic liquid EMIDCA with a 98% concentration is selected as the electrolyte filling in the microchannel for its stability, low volatility, and a relatively high conductivity compared to its counterparts, which are NaCl and KCl solutions, used in previous research [60-62].



Figure 2.2 One sensing segment of the microfluidic device for distributed force sensing (a) side view of the segment (b) deformation of the polymer structure (c) geometry change of the electrolyte and its corresponding resistance change.

The electrodes on one side are connected to a function generator, while the electrodes on

the other side are connected to the circuits that amplify the signals and convert them to DC output. The device essentially has both a capacitive and a resistive nature. The capacitive nature is mainly from the double layer effect formed by the electrode-electrolyte interface. However, with the highly conductive electrolyte EMIDCA, the entire device can be treated as resistive dominant when the passing signal is at high frequencies (>100 kHz), since at high frequencies the capacitor is treated as a wire [61, 63-65]. Besides, DC voltage will cause severe electrolysis and damage the deposited electrodes. In this work, the voltage of the AC signal is kept around 200 mV. As mentioned by Gurierrez et al. [66], low voltage magnitude helps prevent the hydrolysis of the electrolyte. Each opposing pair of electrodes can record the voltage on the electrolyte portion at its corresponding location. Deflection of the PDMS structure causes a corresponding resistance change of the electrolyte in the microchannel and is recorded as a change in the output DC voltage. A device characterization has to be carried out for obtaining the relation between the voltage output and the mechanical input before conducting any measurement on material samples. Once this relation is established, the deformation of the PDMS microstructure at the locations of the transducers can be deduced from the change in the voltage output. A simultaneous spatial measurement thus becomes feasible.

The mechanical input is applied by a rigid probe. Normally, two types of probes are utilized, a cylindrical probe and a probe with a rectangular shape, as can be seen in Figure 2.3. The two types of probes have their own advantages and disadvantages. The cylindrical probe provides a regional deformation above the microchannel, leaving the rest of the device or the rest region of the material, under test minimally affected by the applied load. At the same time, due to the cylindrical shape, misalignment of the probe about its own axis is alleviated. The probe with a rectangular shape covers a larger region of the device, which will give rise to a larger deformation of the microchannel and lead to a larger sensitivity. However, to manually align the rectangular probe parallel to the sensor surface could be difficult.







(b)

Figure 2.3 Photos of (a) a cylindrical probe [60] and (b) a rectangular probe.

2.1.1 Theoretical models

From a mechanical perspective, the force exerted on the polymer structure causes deflection, as shown in Figure 2.2 (b). The force and deflection relation can be expressed using a linear expression:

$$F = K \cdot z \,, \tag{2.1}$$

where K represents the stiffness of the device, and z represents the displacement applied by a probe

on the top surface of the polymer microstructure. In the later analysis, z is also referred to as the indentation depth.

From an electrical perspective, the circuits include a transimpedance amplifier and a demodulation stage. These two stages amplify the signal and convert the signal into a DC output, which can be expressed as:

$$V_{out} = \frac{V_{PP}^2 R_F^2}{8R_i^2(z)}.$$
 (2.2)

In the equation above, V_{PP} denotes the peak-to-peak value of the AC voltage that is applied to the device; R_F is the feedback resistor with a known resistance value, which is 25 k Ω in this work. $R_i(z)$ is the resistance value of the *i*th sensing segment. The schematic of the signal flow is presented in Figure 2.4. The circuit diagram and the derivation of Equation 2.2 are presented in the previous work [60-62] and are also shown in Appendix D.



Figure 2.4 The circuit configuration and signal flow of the microfluidic device for obtaining a DC voltage output.

In Equation 2.2, $R_i(z)$ is a function of the deformation of the polymer structure, which can be further written into:

$$R_{i}(z) = \int_{-\frac{w_{c}}{2}}^{\frac{w_{c}}{2}} \frac{\rho_{E}}{d_{eff}} \cdot h(x, z) dx, \qquad (2.3)$$

where ρ_E is the electrical resistivity of 1-ethyl-3-methylimidazolium dicyanamide, w_C is the microchannel width, d_{eff} is the effective width of each sensing segment or transducer spacing, and h is the local height of the microchannel under external load. ρ_E is 0.37 $\Omega \cdot m$ in Equation 2.3, converted from its conductivity, which is 27 mS/cm [67]; d_{eff} is 0.75 mm for a small device with a smaller transducer spacing and 1.5 mm for a larger device with a larger transducer spacing. h(x,z) is the height of the microchannel, which is a function of the location along the microchannel length direction x, and a function of the deformation of the polymer structure associated with the displacement z applied to the upper surface of the microstructure by a rigid probe.

The integral form can be simplified into [60]:

$$R_i(z) = \frac{\rho_E \cdot w_C}{d_{eff} \cdot h_C} \cdot \frac{1}{1 - \frac{z}{h_c} \cdot \eta}, \qquad (2.4)$$

where h_c is the microchannel height, *z* is the displacement applied to the top surface of the polymer structure by the probe, or the indentation depth. η is a variable that is used to account for the change in the microchannel height. The other symbols are introduced in Equation 2.3. η is not a fixed value, therefore the finite element analysis is employed, for obtaining the relation between R_i and *z*, as will be discussed in the next section.

2.1.2 Finite element analysis

In this section, a simulation based on the finite element analysis method is performed for two microfluidic devices with different key design parameters, which are employed for conducting the CSM of a material in this work. The two devices are of the same design, but have different dimensions. In particular, they have a different PDMS width, PDMS length, PDMS thickness, microchannel width, and transducer spacing. The key design parameters of the two devices are listed in Table 2.1 for comparison.

For simplicity a 2D model is built in COMSOL Multiphysics 4.4 for each device to study the relations between the reaction force and the indentation depth, the resistance of each sensing segment and the indentation depth, and the voltage output of one sensing segment and the indentation depth. Plane strain is assumed in this simulation. Since the device is symmetric, only half of the device is included in the 2D model. Other than the main geometric dimensions of the two devices listed in Table 2.1, the material properties of the devices are shown in Table 2.2.

Symbol	Valua	Unit	Notos
Symbol	value	Unit	INDIES
E_{PDMS}	700	kPa	Young's modulus of PDMS
V_PDMS	0.45	-	Poisson's ratio of PDMS
$ ho_{PDMS}$	1000	kg/m ³	Density of PDMS
E_probe	530	GPa	Young's modulus of the probe
V_probe	0.31	-	Poisson's ratio of the probe
$ ho$ _probe	1560	kg/m ³	Density of the probe
E_pyrex	67	GPa	Young's modulus of the substrate
V_pyrex	0.2	-	Poisson's ratio of the substrate
$\rho_{_pyrex}$	2230	kg/m ³	Density of the substrate

Table 2.2 Material properties used in the simulation.

As can be seen in Table 2.1, the two devices used in this work mainly differ in the microchannel width, the height of the PDMS structure, and the transducer spacing. The microchannel height is only 80 μ m, less than 10% of the total thickness of either of the devices. It should be noted that the model only includes the solid structure, since the amount of the conductive liquid is small enough to be considered to have a negligible reaction force to the microstructure deformation. The two models are built in COMSOL Multiphysics 4.4. For the large device (device

#1), 2014 mesh elements are created, while for the small one (device #2) 2921 mesh elements are created.

Figure 2.5 shows the simulation results for both devices. The displacement fields in the y direction (along microchannel height) for both devices are presented in Figure 2.5 (a) and (b). Although the deformation mainly happens near the rigid probe, the microchannel shape still changes and is able to generate a resistance change in the sensing segment. Figure 2.6 shows the relation between the deformation of the microchannel top and the indentation depth. The displacement along the y-direction of points A and B, as can be seen in Figure 2.5 (a) and (b), represents the displacement applied by the rigid probe on devices #1 and #2, respectively. The displacement along the y direction of points A' and B' represents the deformation of the microchannel top of the two devices.

Figure 2.6 shows the deformation of the top of the microchannel of both devices under a prescribed displacement up to 300 µm applied by a rigid cylindrical probe. The deformation of the microchannel is represented by the displacement of the point located right in the middle of the microchannel top (A' and B' in Figure 2.5). It can be seen that the microchannel of the small device (device #2) deforms less than the large one (device #1), due to its smaller microchannel width and larger PDMS structure thickness.





Figure 2.5 Simulation result of the displacement field at the y direction for (a) device #1 with a thinner PDMS structure and (b) device #2 with a thicker PDMS structure (A and B represent the displacement applied by the probe, and A' and B' represent the microchannel top deformation.).



Figure 2.6 Simulation result of the deformation of the microchannel top under different indentation depth applied by a rigid probe for both devices.

Figure 2.7 shows the relation between the reaction force and the indentation depth of the two devices. The simulation result from the 2D model indicates that with the same cylindrical probe, the stiffnesses of both devices are very close to each other, regardless of the differences in the microchannel width and the thickness of the PDMS structure. From the simulation, the stiffness of the two devices are both around 2500 N/m.



Figure 2.7 Simulation result of the relation between the reaction force and the indentation depth under a rigid probe for both devices.

With the known electrical resistivity $0.37 \ \Omega \cdot m$ of the electrolyte, the resistance of each sensing segment is calculated for different indentation depths according to Equation 2.3, under the assumption that no electrical coupling exists between the segments adjacent to each other. In other words, the entire device is treated as five discrete resistive sensing segments. As can be seen in Figure 2.8, the resistance of each sensing segment changes with the indentation depth applied by the rigid probe. The initial resistance for each segment is 3067 Ω for both devices. Based on the obtained resistance of each sensing segment through FEA analysis and Equation 2.2, the voltage output can be calculated, as can be seen in Figure 2.9. This value is consistent with the experimental data shown in the later section. In Figure 2.9, the voltage output from the small device exhibits a linear relation with the indentation depth. As for the large device, it also can be treated linearly when the indentation depth is small.



Figure 2.8 Simulation result of the relation between the resistance and indentation depth of one sensing segment for both devices.



Figure 2.9 Simulation result of the relation between the voltage output and indentation depth of one sensing segment for both devices.

Figure 2.10 is a 3-D plot that illustrates the microchannel top deformation of one sensing segment, exporting from a full-size 3-D FEA model of the large device (device #1), when the indentation depth is 100 μ m. As can be seen, the width of the sensing segment is 1.5 mm, the length is 1 mm, and the deformation along the x direction is almost uniform, which makes the 2D simplification with the plane strain assumption hold.



Figure 2.10 Deformation of the microchannel top of one sensing segment of the large microfluidic device when the indentation depth is 100 µm.

2.2 Fabrication Process

The main parts of the microfluidic device are the PDMS structure with a microchannel and a piece of Pyrex slide with patterned electrodes. The main steps for fabricating the microfluidic device can be summarized as: photolithography, sputtering, and bonding. Before these main steps, the Pyrex slides should be pre-treated. They are cleaned with a Potassium hydroxide (KOH) solution in an ultrasonic cleaner for 10 minutes, cleaned with acetone for 10 minutes in an ultrasonic bath, rinsed with isopropyl alcohol and DI water later on, and dried with Nitrogen gas.

2.2.1 Photolithography for making SU-8 mold and patterning electrodes

Photolithography is an important step to pattern the electrodes and to make the mold for the microchannel in the PDMS structure. The microchannel shape and electrodes with the desired pattern are first drawn in AutoCad®, as shown in Figure 2.11 (a) and (b). Later on, these patterns are translated to a polyester film as a transparent region, while the rest region of the polyester film remains black coated.



Figure 2.11 Mask design for (a) the microchannel and (b) the patterned electrodes.

To make the mold for the microchannel, a photoresist material SU-8 50 is first spin-coated to a Pyrex slide to form a uniform thin layer. SU-8 is a commonly used epoxy-based negative photoresist that can be used to pattern high aspect ratio structures. It can be spun over a thickness ranging from below 1 μ m to more than 300 μ m before proceeding to photolithography and is now widely used in the fabrication of microfluidic devices [68-70]. After pouring SU-8 on the Pyrex

slide, it is followed by a five-second spin coating at 500 rpm and a 30-second spin coating at 1500 rpm. Then, the slide is baked at 65°C for 10 minutes and 95°C for 30 minutes. After that, the Pyrex slide with an 80 µm thick SU-8 layer is ready for exposure.

The Pyrex slide with SU-8 is then brought to an ultraviolet (UV) light with the microchannel mask mounted on it, allowing the UV light to penetrate the region with the microchannel shape on the mask. Since SU-8 is a negative photoresist material, the part exposed to the UV light is left on the Pyrex slide. In this case, an i-line Karl Suss MJB3 mask aligner is applied to the slide for 15 seconds of exposure. This is followed by a 1 minute post exposure bake (PEB) #1 at 65°C, a 10 minute PEB #2 at 95°C, and a 5-minute development by an SU-8 developer. Then, the Pyrex slide with the SU-8 mold is cleaned with isopropyl alcohol and DI water. After drying the slide, the SU-8 mold with the pattern of an 80 µm thick microchannel is ready to use, as shown in Figure 2.12 (a).



Figure 2.12 Schematics of (a) an SU-8 mold for making the microchannel and two reservoirs on a Pyrex slide, and (b) electrodes patterning with S1800 photoresist on another Pyrex slide.

Similarly, another photoresist material S1800 is used to pattern the electrodes. S1800 is a positive photoresistive material that eventually leaves the part not exposed to UV light on the Pyrex slide. The S1800 layer is also spin-coated on the Pyrex slide, at 500 rpm for 10 seconds and at 2000 rpm for 30 seconds. After being soft baked and hard baked at 90°C for 1 minute, the Pyrex slide covered with the electrodes mask is placed under an Exoteric 405 nm UV Flood Source for 5 seconds at attenuation 3. Then the pattern of electrodes is formed with the help of developer MF24, as can be seen in Figure 2.12 (b).

2.2.2 Fabrication of electrodes and PDMS structure

The Pyrex slide with the S1800 sacrificial layer is treated with oxygen plasma for thorough cleaning and better adhesion for the metal layer to deposit. The sputtering system is EMITECH K675X Turbo Large Chromium Coater. The Pyrex slide is first deposited with Chromium (Cr) of 10 nm thick, and then gold (Au) with 100 nm thickness. Then the entire Pyrex slide is washed with acetone in ultrasonic to remove the photoresist, leaving only the metal electrodes patterned on the slide. After being rinsed in isopropyl alcohol and DI water, the slide is dried with Nitrogen gas. Then the fabrication of the electrodes is finished. Figure 2.13 shows the schematic of patterned electrodes.



Figure 2.13 Au/Cr electrodes on the Pyrex slide by sputtering and metal deposition.

PDMS is used to make the polymer structure for this microfluidic device. A mixture of an elastomer and a curing agent (Sylgard 184kit, Dow Corning Corp.) with a ratio of 10:1 is poured over the SU-8 mold and cured for 24 hours at room temperature on the optical table to ensure an even thickness. The thickness of the PDMS structure can be adjusted by the amount of the mixture being poured on the mold. Figure 2.14 shows the SU-8 mold of the microchannel being covered by the 10:1 PDMS mixture and is cured at room temperature.



Figure 2.14 SU-8 mold of the microchannel being covered by the 10:1 PDMS mixture cured at room temperature.

After sitting at room temperature for 24 hours, the PDMS structure is peeled off from the SU-8 mold, and then treated with oxygen plasma for cleaning. Then, it is bonded to the slide with patterned electrodes under a microscope. In the final step, a 1-Ethyl-3-methylimidazolium dicyanamide (EMIDCA) is then injected in the microchannel through two reservoirs. Figure 2.15 shows the fabricated microfluidic device with electrolyte in the microchannel.



Figure 2.15 Fabricated microfluidic device with electrolyte in the microchannel.

2.3 Performance Characterization

The two microfluidic devices were first experimentally characterized, and the detailed characterization process is elaborated in Chapter 4. A cylindrical probe as shown in Figure 2.3 (a) was assembled to a micromanipulator to apply displacement with a programmed pattern in this characterization. It was manually aligned right above the microchannel, covering all the electrodes and then finely adjusted to ensure a proper alignment, without obvious tilting towards any direction.

Figure 2.16 and Figure 2.17 illustrate the characterization results corresponding to devices #1 and #2, respectively. For device #1, the indentation of the device reaches up to 200 μ m. The stiffness is calculated as 2339.4 N/m from the *F* and *z* relation, as can be seen in Figure 2.16 (a). Figure 2.16 (b) shows the voltage output from the five transducers of the device, as the indentation depth increases. They all show a linearly decreasing trend. Similarly, the characterization result for device #2 is presented in Figure 2.17. The stiffness is calculated as 1849.5 N/m, from the force-indentation depth relation. The relation between the voltage output and the indentation depth also can be linearly fitted, as can be seen in Figure 2.17 (b).



Figure 2.16 Relation between (a) force and indentation depth and (b) voltage output from five transducers and indentation depth of device #1 (large device).



Figure 2.17 Relation between (a) force and indentation depth and (b) voltage output from five transducers and indentation depth of device #2 (small device).

It can be observed that the overall trend of the force-indentation relation and the voltage output-indentation relation obtained from both the experiment and simulation are consistent.

However, in the experimental characterization, the voltage change is smaller than it is in the result obtained by the numerical simulation.

The reasons for the discrepancy can be summarized as follows:

1) The actual dimensions of the devices vary a little from the nominal dimensions, especially for the microchannel. The microchannel height has the smallest dimension among the key design parameters. It plays an important role in the voltage output and is difficult to be accurately measured. The initial resistance of one sensing segment for both devices is supposed to be 3067Ω theoretically, corresponding to a microchannel height of 80 µm. The theoretical voltage output from the circuit is supposed to be 0.332 V, corresponding to a feedback resistance of $25 \text{ k}\Omega$ and 200 mV peak-to-peak voltage input. However, from Figure 2.16 (b) and Figure 2.17 (b), it is noticed that the initial voltage output from the five transducers of both devices is less than 0.332 V, which could be caused by fabrication defects. For instance, a less than 80 µm microchannel height for both devices can cause such situations. The voltage output of the five transducers also varies, which can be attributed to the fabrication variations, especially in the microchannel height at the transducer locations. The effect of the microchannel height on the resistance and voltage output of each sensing segment is discussed in Section 2.4.

2) Mechanical and Electrical Crosstalk. In the previous simulation, it is assumed that the effective width of each sensing segment equals the transducer spacing, and no mechanical or electrical crosstalk exists between the neighboring sensing segments. However, the real interaction between the neighboring sensing segments is more complicated. From the electrical side, the conductive portion is a liquid that flows between the sensing segments, making the interaction between the neighboring transducers inevitable. From the mechanical side, all sensing segments

share one piece of the PDMS structure. Although the low elastic modulus of the PDMS alleviates the coupling, crosstalk from the mechanical side still exists.

3) Assembly issues. The cylindrical probe is mounted on a probe holder in a cantilever shape, which is further assembled to the micromanipulator, as will be seen in Chapter 4. Although the stiffness of the probe is much larger than the device with a PDMS structure, the mechanical assembly may cause the actual displacement applied to the device that is smaller than it is supposed to be, since the prescribed displacement is directly applied by the cantilever tip.

Despite the slight discrepancy between the simulation and the experimental results, based upon the previous analysis, it still can be concluded that a linear relation holds between the voltage output of the device and the indentation depth applied by the cylindrical probe. This conclusion is crucial to the data analysis in the later chapters.

2.4 Technical Issues Encountered

Several technical issues associated with the device itself and the performance characterization are discussed in this section, including probe misalignment, fabrication variation, and the voltage drift problem.

2.4.1 Misalignment of the probe

Misalignment of the probe will bring in an initial non-uniform microchannel top deformation, which further affects the voltage output. There are three types of misalignment: z-direction misalignment, in-plane misalignment, and axial misalignment.

2.4.1.1 z-direction misalignment of the probe

Figure 2.18 (a) schematically presents the z-direction misalignment of the probe. In this case, the probe is not perfectly aligned horizontally and rotates several degrees about the y-axis. The probe tilts towards one end of the microchannel, causing the microchannel top to have non-

uniform deflections. As can be seen in Figure 2.18, the 11 mm long probe tilts towards the right end of the microchannel by 0.1 degree when a 60 μ m displacement is applied by it, caused the non-uniform microchannel top deflections. This type of probe misalignment causes an increasing or decreasing trend in the voltage output.



Figure 2.18 (a) Schematic of the z-direction misalignment of a probe (front view) and (b) simulation result of microchannel top deformation of device #1 under the z-direction misalignment of the probe, when the applied displacement equals to 60 μm.

2.4.1.2 In-plane misalignment of the probe

The in-plane misalignment is depicted in Figure 2.19 (a). In this case, the probe does not perfectly cover the microchannel, and rotates about the z-axis by several degrees, 2.3° in the simulation. Similar to the previous case, the in-plane misalignment of the probe also leads to the microchannel top having non-uniform deformation along the z direction, which will ultimately cause the variations in the voltage output from transducers.



Figure 2.19 (a) Schematic of the in-plane misalignment of the probe (top view) and (b) the effect of the in-plane misalignment on the microchannel top deformation under 60 µm prescribed displacement.

2.4.1.3 Misalignment about the x-axis

As mentioned in Section 2.1, due to the circular shape of a cylindrical probe, the tilt of the probe about the x-axis does not affect the contact between the probe and the device, so that the voltage output will not be affected. The schematic of this type of misalignment is illustrated in Figure 2.20.



Figure 2.20 Schematic of the misalignment about the x-axis (side view).

2.4.2 Fabrication variations

Fabrication variations are other issues that affect the device performance. Microchannel height is one of the variations that could significantly affect the voltage output from the device, since it has the smallest dimension among all the key design parameters. The microchannel height of the device is determined by the thickness of the SU-8 mold used for fabricating the microchannel, as mentioned in Section 2.2.1. It is inevitable that the thickness of the mold has some variations. A simulation of the sensor with a varying microchannel height is carried out. The microchannel height varies from 50 μ m to 80 μ m, from the left to the right, as shown in Figure 2.21. As can be seen in Table 2.3, the initial resistance decreases from transducer 1 to transducer 5. When a 60 μ m prescribed displacement is applied, the corresponding resistance and voltage output values of each sensing segment are also listed in Table 2.3.



Figure 2.21 Schematic of a device with variations in microchannel height (front view).

Table 2.3 Resistance and voltage output of a device with varying microchannel height at the locations of the transducers.

Transducer	1	2	3	4	5
Initial resistance (Ω)	3982	3888	3798	3713	3631
60 μ m resistance (Ω)	3996	3901	3811	3726	3642
Initial voltage (V)	0.1971	0.2067	0.2166	0.2267	0.2370
60 µm voltage (V)	0.1957	0.2054	0.2152	0.2251	0.2356
$\Delta V(\mathbf{V})$	0.0014	0.0013	0.0014	0.0016	0.0014

2.4.3 Voltage drift

So far, NaCl, KCl solutions and a 1-ethyl-3-methylimidazolium dicyanamide have been used as the electrolyte for the microfluidic device. It is proved in a previous publication that a 1ethyl-3-methylimidazolium dicyanamide has very low volatility, compared with NaCl, KCl solutions, and is desirable for being a stable conductive liquid [62]. Despite the excellent electrical stability, a voltage drift is still detected from the experiment.

Figures 2.22 and 2.23 exhibit the voltage drift of the device during one week and during one dynamic measurement which lasts for one to two hours. To investigate if the voltage output would be stable after the heat equilibrium is established, the device is connected to the circuit and left on for 48 hours. The test result is shown in Appendix B. Overall, the voltage output shows a decreasing trend with time. It means if the voltage drift is caused by the heat dissipation, 48 hours are not enough to allow the device to reach its heat equilibrium, which is not practical in the real operation. Other groups have also reported on the voltage drift phenomenon during the operation of a microfluidic device. Chossat *et al.* [71] attributed the drift to the evaporation of the water in the electrolyte, eutectic gallium indium (eGaIn), due to an imperfect bonding, sealing, or the permeability of the material. Collins et al. [72] also mentioned the voltage drift in their work which employed sodium hydroxide (NaOH) solution as the electrolyte. Its ability of absorbing CO₂ in the air and form carbonic acid is accounting for the voltage drift. Muth et al. [73] used carbon conductive grease as the conductive liquid in their 3D printed devices, but the voltage drift problem still exists. Scott et al. [74] mentioned Ag/AgCl electrodes are electrochemically stable, which can avoid the artifacts from voltage drift caused by polarizable metal electrodes. To the best knowledge of the author, voltage drift issues in microfluidic devices have not been thoroughly studied or solved yet. This issue is reported in this section for completeness.



Figure 2.22 Initial voltages of the device output measured within one week.



Figure 2.23 Voltage output of the device at its vibration equilibrium point measured during one dynamic test.

2.4.4 Discussion

In this section, technical issues associated with the device fabrication and the performance

characterization are discussed. To address the voltage output variations caused by the probe misalignment or fabrication variations, instead of using the direct voltage output in the data analysis, the voltage change from each transducer is used in the later calculation. From both the simulation and the performance characterization, linear relations are observed in the voltage output and indentation relations. Thus, despite the inconsistency in the initial voltage values caused by the misalignment of probe or fabrication variations, the device is still able to detect the change in the input. As a result, the voltage change from each transducer and its slope should be used in the later calculation. Meanwhile, each device employed in the CSM should be characterized as the control experiment before use to account for the inevitable device-to-device variations.

However, the voltage drift issue still requires further investigation. Compensation methods for alleviating voltage drift problem have been explored in other types of sensors [75], the same concept could be adopted and developed in the future for the microfluidic type devices.

2.5 Conclusions

To avoid unnecessary complexity, the current design of the microfluidic device for distributed-load detection comprises of five sensing segments for distributed-load detection. The theoretical models for the microfluidic device from both the mechanical and electrical perspectives are presented, which are the basis for the further analysis on the CSM technique with the incorporation of the device. An effective finite element 2D model is also built to simulate the sensor response from both the mechanical and electrical perspectives and is compared with the results obtained from the performance characterization. Based upon the theoretical and simulation results, the relation between the prescribed displacement (indentation depth) and the device output is built. A linear relation between the voltage change from each transducer and the indentation

depth will be used in the data analysis for extracting the viscoelastic properties of testing materials in later chapters. The fabrication process for the polymer-based microfluidic device shows its simplicity and low cost. Several technical issues associated with the device itself and the testing method are also discussed to account for the potential errors the output data from the device could carry.

The performance of the device is robust, however from the design and fabrication perspectives, there are several aspects worth being improved in a future design. For example, the mechanical and electrical interference between the sensing segments should be considered and alleviated in a future design. Other more compliant elastomers can be used to build the mechanical structure of the microfluidic device for larger deformation or stretching. Other conductive liquids could be employed as the electrolyte to alleviate the voltage drift problem existing in the current sensor. Meanwhile, other fabrication techniques, such as 3D printing, could also be considered in the future.

Overall, in this chapter, the polymer-based microfluidic device employed for the CSM in later chapters is presented. The working principle, fabrication process, performance characterization, and technical issues encountered in the experimental characterization are thoroughly discussed.

CHAPTER 3

CONCURRENT SPATIAL MAPPING OF THE ELASTIC MODULUS OF SOFT MATERIALS: THEORY

This chapter explores the theory and rationale of conducting the CSM technique to obtain the spatially-varying elastic modulus of heterogeneous soft materials via the microfluidic device. This chapter first describes the related knowledge in the field of contact mechanics associated with the CSM technique. Then the rationale of conducting the CSM is explained in detail. Next, the capability of detecting the heterogeneity in a sample through the microfluidic device is studied through the FEA method. The effect of the heterogeneity parameters on the device output is also explored.

3.1 Contact Mechanics

In a CSM of the elastic modulus, a cylindrical probe is used to press a material sample against the device to obtain the force and indentation depth relation at the locations of the transducers. Thus, it is essential to obtain an accurate model to relate the elastic modulus to the measurable force and indentation depth applied to the material sample.

Figure 3.1 schematically shows a cylinder probe is in contact with a flat sheet with infinite thickness and indents the flat sheet with a depth of δ . Puttock and Thwaite [2] derived the relation between the reaction force and the indentation depth of a cylinder in contact with a flat sheet with infinite thickness. As can be seen in Equation 3.1, the indentation depth δ is related to the reaction force *F* and the material properties.



Figure 3.1 Schematic of a flat sheet under the indentation of a cylinder.

$$\delta = F / (2a) * (V_1 + V_2) \cdot \left[1 + \ln \left(\frac{16a^3}{(V_1 + V_2) \cdot F \cdot D} \right) \right]$$
(3.1)

$$V = (1 - v^2) / \pi E \tag{3.2}$$

In Equation 3.1, *a* denotes the half-length of the cylinder; *D* denotes the diameter of the cylinder. Equation 3.2 shows the expression for V_1 and V_2 , which are related to the elastic modulus *E* and Poisson's ratio *v* of two materials that are touching.

However, the equations above are based on the assumption that the material under indentation has infinite thickness. A finite element model is built in COMSOL, to verify the relation between the reaction force and the indentation depth exerted by a cylinder probe, as can be seen in Figure 3.2. The model contains 197,529 elements, with an average element quality of 0.7539. The material is assumed to be elastic in the simulation. As proven by Biersack [76], the results obtained by assuming the material having hyperelastic property is very close to the results as a linear elastic property is assumed.


disp(6)=3e-5 Surface: Displacement field, Z component (m)

Figure 3.2 A finite element model for verifying the relation between the reaction force and the indentation depth exerted by a cylinder.

In the simulation, the material under the indentation of the cylinder probe has the dimension of 20 mm \times 10 mm \times 10.5 mm, with the elastic modulus of 700 kPa, Poisson's ratio of 0.45, and density of 1000 kg/m³. The probe has an 11 mm length, a 0.8 mm radius, and has the material property of 530 GPa for the elastic modulus, 0.33 for the Poisson's ratio, and 2.23 kg/m³ for the density. The indentation depth is from 0 to 30 µm, which is much smaller comparing with the thickness of the material under indentation. The probe radius is also much smaller than the thickness of the material, which makes it reasonable to assume the thickness of the material under indentation model.

Figure 3.3 shows the theoretical and simulation results of the contact model between a cylinder and a flat sheet with infinite thickness. It can be seen that the simulation result obtained in COMSOL Multiphysics agrees very well with the theoretical model.



Figure 3.3 Theoretical and FEA simulation results of the contact model of a cylinder and a flat sheet with infinite thickness.

However, the thickness of a sample will start to affect the accuracy of the results obtained by Equations 3.1 and 3.2, when the thickness of the sample gets close to the diameter of the probe or the indentation depth. It is necessary to bring in a correction factor to Equation 3.1 in order to achieve a more accurate elastic modulus that relates the force and indentation depth. For a spherical indenter that is commonly used in AFM or a nanoindenter, modifications were made to the widely used Hertz model to account for the effect of the finite thickness of a sample by Dimitriadis [25]. However, to author's knowledge, there is sparse research that highlights the derivation of the analytical form of a cylinder in contact with a flat sheet with finite thickness, which only applies to special cases [77-79]. There is no analytical solution in a general case to account for the effect of the thickness of a flat sheet on the overall force and indentation relation under the press of a cylinder probe. To address this dilemma, the same simulation method, as mentioned above, is applied to models in which the thickness of the flat sheet is small. The force and indentation relations are achieved for sheets with different thicknesses, and are presented in Figure 3.4.



Figure 3.4 Simulation results of force-indentation depth relation with different sheet thicknesses.

From Figure 3.4, it can be seen that the thickness of the sheet affects the force and indentation relation. When the thickness of the sample is small, the elastic modulus E' obtained by curve fitting Equations 3.1 and 3.2 becomes larger than it is supposed to be. Thus, a correction term α should be added to obtain the true elastic modulus of the materials.

$$E = \alpha \cdot E' \tag{3.3}$$

For convenience, the correction factor is obtained by comparing the extracted elastic modulus of the materials with different thicknesses from Equation 3.1 and the true elastic modulus

defined in the simulation. The relation between the correction factor α and the thickness is plotted in Figure 3.5. The expression of the correction factor can be written as

$$\alpha = 5.12h^{0.32}, \tag{3.4}$$

where h is the thickness of the sample. Equations 3.3 and 3.4 will be later used in the data analysis.



Figure 3.5 The relation between the correction factor α and the thickness of a flat sheet.

Since the probe has a very large elastic modulus, it can be considered as rigid. With the correction factor, Equation 3.1 can be further simplified to

$$\delta = \frac{5.12h^{0.32}(1-\upsilon^2)F}{2a\pi E} \cdot \left[1 + \ln\left(\frac{16\pi Ea^3}{5.12h^{0.32}(1-\upsilon^2)\cdot F\cdot D}\right) \right].$$
 (3.5)

The correction factor is obtained based on the models with a thickness range of 1 mm to 3.5 mm in COMSOL Multiphysics, which is consistent with the real application in this work. Thus,

the best application range of Equation 3.5 will be when the thickness of the flat sheet is also around 1 mm to 3.5 mm.

3.2 Rationale

In this section, the rationale of the CSM technique for the elastic modulus measurement is presented.

Figure 3.6 schematically shows the side view of the device with one piece of test material placed on top of it, pressed by a cylindrical probe against the device. In the measurement, one piece of soft material with unknown elastic modulus is put between the device and the probe. As can be seen in Figure 3.6, the total displacement z, or the total indentation depth, applied by the rigid probe can be expressed by the summation of z_s and z_d , which stand for the deformations of the sample and the device, respectively.

$$z = z_s + z_d \tag{3.6}$$



Figure 3.6 Schematic of the side view of one piece of soft material under indentation applied by a cylindrical probe.

Figure 3.7 depicts the rationale of the CSM technique for the elastic modulus measurement of a piece of soft material through the polymer-based microfluidic device. One piece of material with an unknown elastic modulus is placed on the device, and a rigid cylindrical probe is employed to apply a precisely-controlled displacement to press the material against the device. There are two outputs from the elastic modulus measurement from the CSM technique. One is the overall load, which is read from a load cell assembled in the setup, and the other is the voltage output from the device, representing the device deflection.



Figure 3.7 Rationale of the CSM technique for the elastic modulus measurement.

$$F_d = \sum K_{di} \cdot z_{di} \tag{3.7}$$

$$R_{i}(z) = \int_{-\frac{w_{c}}{2}}^{\frac{w_{c}}{2}} \frac{\rho_{E}}{d_{eff}} \cdot h(x, z) dx$$
(3.8)

$$V_i = \frac{V_{PP}^2 R_F^2}{8R_i^2(z)}$$
(3.9)

In Equation 3.7, the overall load applied to the material is measured by the load cell as output one. Equations 3.8 and 3.9 illustrate the resistance of each sensing segment and the voltage output from each sensing segment as output two.

Ideally, when a homogeneous sample is being pressed against the device by the cylindrical probe mentioned in the previous chapter, a uniform deformation of the sample and the device will be generated, leading to the same amount of change in the resistance of the five sensing segments. As a result, the output from the five transducers should be the same. In Figure 3.7, the sample under test is a heterogeneous material, which causes a spatially-varying elastic modulus distribution of the sample. Different from the homogeneous sample, the stress and strain distribution will be changed by the heterogeneity, leading to a continuous non-uniform deflection of the device, which further causes the differences in the resistance change and the voltage output. As can be seen in Figure 3.7, the resistance change of transducer four is different from the others, due to the existence of the heterogeneity composition with a larger elastic modulus at its corresponding location in the sample.

The implementation of the CSM of the spatially-varying elastic modulus of a piece of soft material is elaborated in Chapter 4.

3.3 Finite Element Analysis

The capability of detecting the heterogeneity within a piece of soft material via the microfluidic device is verified in this section through an FEA approach. A finite element model is built in COMSOL Multiphysics. The size, geometry, and material properties of the device are the same as device # 1 mentioned in Chapter 2. The dimension and the material property of the probe being used to press the sample are also the same as the one mentioned previously in Chapter 2. A piece of homogeneous soft material with an elastic modulus of 200 kPa, Poisson's ratio of 0.45

and density of 1 kg/m^3 is first built between the probe and the device. The in-plane dimension of the piece of material is $15 \text{ mm} \times 6 \text{ mm}$; the thickness is 2.5 mm. It is placed on top of the device, covering the five transducers. A piece of heterogeneous soft material is modeled in the second simulation. To simplify the model, a sphere-shaped geometry with a radius of 0.5 mm and a high elastic modulus of 2 GPa is built in the model to mimic the heterogeneity inside a material. As can be seen in Figure 3.8, the hard sphere is located above one of the transducers.



Figure 3.8 Simulation of a piece of heterogeneous material under the indentation applied by a cylindrical probe on a microfluidic force sensor.

Figure 3.9 presents the simulation result of the microchannel top deformation when a piece of homogeneous material is being pressed against the device. The microchannel is 30 mm in length, and the probe is 11 mm long, which explains the majority of the deformation happens in the central area of the microchannel top. The five transducers are 1.5 mm away from each other, locating in the middle section of the microchannel, along the microchannel length. To be specific, they are located at 12 mm, 13.5 mm, 15 mm, 16.5 mm and 18 mm, corresponding to the horizontal

coordinates in Figure 3.9 and Figure 3.10. The total indentation depth applied by the probe for both simulations is 100 μ m. Under this prescribed indentation depth, the simulation result obtained for the homogeneous sample exhibits consistent deformation at the locations of the transducers, with very slight variations, as can be seen in Figure 3.9. The simulation result obtained from the heterogeneous model exhibits sudden changes in the microchannel top deformation, when the heterogeneity is located above the corresponding transducer, as can be seen in Figure 3.10. Thus, the location of the heterogeneity is able to be detected by identifying an abrupt change in the output signal from the five transducers.



Figure 3.9 Simulated microchannel top deformation when a piece of homogeneous material is under the indentation applied by a cylinder probe on the microfluidic device (The dashed lines represent the locations of the transducers).



Figure 3.10 Simulated microchannel top deformation when a piece of heterogeneous material under compression by a cylinder probe on the microfluidic force sensor. (The dashed lines represent the locations of the transducers.)



Figure 3.11 Simulation of a piece of material sample with a hard sphere (r = 1.2 mm) embedded, indented by a cylindrical probe on the microfluidic device.

The effect of the depth (d_h), size (r_h), and elastic modulus (E_h) of the heterogeneity inside a material on the microchannel top deformation is also investigated. To keep consistency, the hard sphere is located above the middle transducer in all the simulation models, the applied indentation depth is 100 µm in all the models. In the first model, the distance between the center of the embedded hard sphere and the device surface varies from 0.5 mm to 1.7 mm. Figure 3.12 presents the deformation of the microchannel top along the microchannel length, as the embedded sphere locates further away from the device. As it is expected, right above the middle transducer, the deformation of the microchannel top is larger than the other locations. However, a more abrupt change happens when the hard embedded sphere is closer to the device. The deformation tends to be overlapping each other, when the distance between the sphere and the device reaches 1.3 mm.



Figure 3.12 Deformation of the microchannel top along the microchannel length, when the distance between the embedded sphere and the device surface increases (The dashed lines represent the locations of the transducers).

In the second simulation model, the radius of the embedded sphere varies from 0.2 mm to 1.2 mm with a 0.2 mm interval, while the center of the hard sphere is located right in the middle of the sample. Figure 3.13 shows the deformation of the microchannel top along the microchannel length, as the radius of the embedded sphere increases. As can be seen, the hard sphere is no longer detectable, when the radius is below 0.4 mm. It can be concluded that the radius of the hard sphere has a larger impact on the microchannel deformation, since an approximate 40 µm change occurs at the microchannel top, when the sphere radius is as large as 1.2 mm.



Figure 3.13 Deformation of the microchannel top along the microchannel length, when the radius of the embedded sphere increases (The dashed lines represent the locations of the transducers).

In the third model, the varying parameter is the elastic modulus of the sphere, which varies from 500 kPa to 4.9 MPa. The sphere representing the heterogeneity has a radius of 0.5 mm, and is located in the middle of the sample. The simulation result is presented in Figure 3.14. It is noticed that although the deformation of the microchannel top increases with the elastic modulus of the

sphere, the change of the deformation is small, and is even smaller after the value of the elastic modulus of the sphere reaches 2.5 MPa.

The results indicate that the microchannel deformation is most sensitive to the size of the heterogeneity inside a sample. The embedded depth and the elastic modulus of the heterogeneity also make noticeable differences in the output, but not as much as the size's effects on the output.



Figure 3.14 Deformation of the microchannel top along the microchannel length, when the Young's modulus of the embedded sphere increases (The dashed lines represent the locations of the transducers).

3.4 Discussion and Conclusions

In this chapter, the rationale of the CSM technique for the elastic modulus measurement through a microfluidic force sensor is elaborated. The model of extracting the elastic property of a piece of material from its force and indentation depth relation is proposed, when the material is of limited thickness and is indented by a cylindrical probe. The capability of detecting the heterogeneity within a piece of material is also discussed and verified through an FEA simulation. From sweeping three parameters, the depth, radius, and elastic modulus of the hard sphere representing the heterogeneity in the FEA model, it can be concluded that the size of the heterogeneity affects the microchannel top deformation the most, which further leads to a larger change in the voltage output from the corresponding transducer of the device.

CHAPTER 4

CONCURRENT SPATIAL MAPPING OF THE ELASTIC MODULUS OF SOFT MATERIALS: IMPLEMENTATION

This chapter focuses on the implementation of the CSM technique for the elastic modulus measurement of both homogeneous and heterogeneous soft materials via the polymer-based microfluidic device. The experimental setup, sample preparation, and experimental protocol are presented in detail. The data analysis method and results are also presented.

4.1 Materials and Methods

The experimental setup, sample preparation method, and the experiment process are elaborated in this section.

4.1.1 Experimental setup

The experimental setup for implementing the CSM technique on a material sample for extracting its elastic modulus is shown in Figure 4.1. A device is mounted on a PCB board, which is further fixed on the optical table through an NBM513 - NanoBlock 5-Axis Waveguide Manipulator (Thorlabs). Materials are cut into rectangular pieces and placed on the device one piece at a time. Mounted on another micromanipulator (MP-285, Sutter Instrument), a probe holder is utilized to assemble a load cell and a rigid probe together. The micromanipulator allows moving the probe along the z-axis to exert precisely-controlled displacement (indentation depth) on a sample. Figure 4.2 shows a photo of the microfluidic force sensor employed for the elastic modulus measurement in this work. As can be seen, a container made by PDMS with an agent-elastomer mixing ratio of 1:10 is built upon this device for accommodating samples that need to be tested in an aqueous condition.



Figure 4.1 Experimental setup for conducting concurrent spatial mapping.



Figure 4.2 A polymer-based microfluidic device with 30 mm microchannel length and a PDMS container to accommodate materials needing to be tested in an aqueous condition [80].

4.1.2 Sample preparation

Both homogeneous and heterogeneous agar samples with an agarose concentration of 1%, 2%, and 3% were prepared in phosphate buffered saline (PBS) from Sigma Aldrich. The solution was heated to boil and was poured into a Petri dish placed on an optical table and allowed to cool. The samples were then cut into small pieces and placed in PBS.

Heterogeneous agar samples with the same concentrations were also prepared, with corn crumbles imbedded. The corn crumbles were about the size of 1 to 2 millimeters, with different shapes and locations in the agar samples. For preparing the heterogeneous samples, corn crumbles were placed in the three Petri dishes first, and then solutions of different concentrations were poured over the crumbles within their respective dishes.

The solidification time for preparing both types of samples was over an hour at room temperature. Then small pieces of agar samples were cut out with an in-plane dimension of approximate 10 mm \times 6 mm. The thickness of the samples was then measured. The samples were then stored in PBS at room temperature until being tested.

Similar to the preparation of the agar samples, both homogeneous and heterogeneous PDMS samples were made, as shown in Figure 4.3. The curing agent and the elastomer with the ratio of 1:30 and 1:40 were poured into two different containers and then thoroughly mixed. For preparing the homogeneous PDMS samples, the elastomer-agent mixtures with two different mixing ratios were separately poured into two flat containers being placed on an optical table. For preparing heterogeneous samples, the mixtures were poured into containers with those scattering raw corn crumbles mentioned above. The containers were placed on the optical table for over 24 hours to cure before use. Owing to the volume of the mixture being used, the thickness of samples

varies from one another. A sample thickness measurement is required for data analysis, and all the sample thicknesses are presented in Table 4.1.



Figure 4.3 Prepared (a) 1:30 homogeneous PDMS (b) 1:30 heterogeneous PDMS and (c) 1:40 heterogeneous PDMS samples [80].

Sample	Thickness			
	(mm)			
1% homogeneous agar	1.954			
1% heterogeneous agar	2.638			
2% homogeneous agar	2.253			
2% heterogeneous agar	2.253			
3% homogeneous agar	2.502			
3% heterogeneous agar	2.264			
1:30 homogeneous PDMS	3.558			
1:30 heterogeneous PDMS	3.22			
1:40 heterogeneous PDMS	2.849			

Table 4.1 Thickness of the prepared samples.

4.1.3 Experimental protocol

Prior to conducting the CSM on a piece of material to obtain its elastic modulus, a control experiment on the device is required to acquire its static performance. Specifically, the relation between the precisely controlled displacement (indentation depth) applied by the cylindrical probe to the device upper surface and the voltage output from the transducers should be established.

a) Probe alignment

Probe alignment should be conducted first to minimize errors caused by misalignment in the output, as discussed in Section 2.4.1. This process is completed with the help of a five-axis manipulator where the device sits, a micromanipulator, and a LabVIEW program that displays the output. The probe is first visually placed above the device, parallel to the microchannel in the device. Then the probe is moved down by the micromanipulator to touch the device. From the LabVIEW interface, the five voltage outputs can be observed simultaneously. Through the 5-axis manipulator, the orientation of the device can be adjusted to ensure the change of the five voltage outputs happens at the same time, which indicates a proper alignment.

b) Control experiment

Prior to measuring a piece of soft material via the polymer-based device, performance characterization of the device should be conducted as a control experiment, for obtaining the relation between the applied displacement *z* (indentation depth) and the voltage output from the five transducers u_i . With the $\Delta u_i - z$ relation, the deflection of the device can be calculated according to Δu_i in future tests.

The control experiment is carried out right after the alignment is complete. The probe is moved by the micromanipulator with a programmed pattern. Figure 4.4 shows one of the patterns used to characterize the device. In the characterization process, the probe is first brought down by the micromanipulator by 20 μ m at the speed of 1 mm/s, stays still for 5 seconds, then goes back to its original location at the same speed. This process is repeated 20 to 40 times, and each time the probe is brought down 20 μ m further than the previous time.



Figure 4.4 Displacement pattern of the probe for conducting device characterization.

c) CSM for the elastic modulus measurement

After the control experiment is completed, the probe is brought up by the micromanipulator, to make space for inserting a piece of prepared sample between the device and the probe. After the material sample is inserted, the probe is brought closer to the sample by the micromanipulator to identify the contact point. Similar to the alignment process, the contact point is identified by observing the voltage output from the transducers. The probe is first brought closer to the sample by the micromanipulator and stops when the distance is closer than 1 mm, judging by visual examination. Then the probe is brought down by 20 μ m each time until an obvious change in voltage output is observed. The probe is then brought up by 10 μ m, and that point is chosen to be the initial point that the probe and sample start to contact.

The CSM is conducted based on the contact point. The motion pattern of the probe used for the CSM is similar to the pattern that is shown in Figure 4.4. The only difference is the indentation depth reaches further than testing the device alone.

For PDMS samples, the largest indentation depth applied by the probe is 500 μ m, while for agar samples, it is 300 μ m. For each sample, the CSM test is applied for 3 times.

4.2 Data Analysis

Equation 3.5 is used to relate the force and indentation depth applied to a sample. The force and indentation depth applied to the sample can be obtained through the CSM measurement. From these two sets of data and the measured thickness of the sample, the elastic modulus of the sample can be calculated. A flow chart of the data processing procedure is presented in Figure 4.5 for better illustration.



Figure 4.5 Flow chart of the data processing procedure of the CSM of the elastic modulus of soft materials.

As mentioned in Section 3.2, the first direct output of the measurement is the overall load applied to the sample, which is calculated from the voltage output of the load cell, through a linear relation. The second output from the measurement is the voltage signals from the five transducers of the device, which represent the deformation of the device at five discrete locations along the microchannel length. The overall indentation is predefined as an input value in the LabVIEW program. According to the conclusion from Chapter 2, the voltage output u_i from each transducer of the device and the indentation applied to the device, z, are linearly related. Thus, the relation between the indentation depth z_{di} and the voltage change Δu_i can be expressed by Equation 4.1.

$$z_{di} = a\Delta u_i + b \tag{4.1}$$

Then the deformation of the device can be acquired by multiplying a constant coefficient a to the voltage change from its corresponding transducer and adding an intercept b. The distributed deformation of the device at the locations of the transducers, z_{di} , is obtained. The deflection of the material sample at the locations of the five transducers can be calculated accordingly by Equation 3.6 from the last chapter, $z_{si} = z - z_{di}$. Meanwhile, Equation 3.5 can be further written as

$$z_{si} = \frac{5.12h^{0.32}(1-\upsilon^2)F}{2a\pi E} \cdot \left[1 + \ln\left(\frac{16\pi Ea^3}{5.12h^{0.32}(1-\upsilon^2)\cdot F\cdot D}\right)\right].$$
 (4.2)

A numerical method is applied to solve Equation 4.2, for obtaining the elastic modulus E.

4.3 Results

4.3.1 Device characterization results

From the control experiment, the coefficients to relate the voltage change from each transducer and the indentation depth as expressed in Equation 4.1 are obtained by a linear curve fitting and are presented in Table 4.2. The negative sign of coefficient a in Table 4.2 indicates that

the voltage output decreases when the indentation depth goes up. Coefficient b is the intercept of the curve fitting, which is very close to 0. It can be seen that among transducers, these coefficients have slight variations, caused by fabrication variations and the misalignment of the probe. Later on, these coefficients will be used to deduce the deformation of the device at the locations of the transducers, based on the voltage change of the output from the transducers.

Table 4.2 Coefficients that relate the indentation depth and the change of voltage output from the device.

Transducers	a (µm/V)	b (um)
1et	-2511.85	2 71
	-2311.03	2.71
2nd	-2447.89	-1.33
3rd	-2132.59	-0.38
4th	-2612.94	0.63
5th	-2499.74	-1.91

4.3.2 Measurement results

In this section, curve fitting for Equation 4.2 is carried out. For PDMS samples, the elastic modulus of each sample is obtained from the overall force-indentation depth curve obtained from the average of the three-time CSM measurements, assuming the elastic modulus is a constant value. However, for agar samples, the obtained elastic modulus results indicate an indentation depth-dependent trend.

4.3.2.1 Measurement results for PDMS samples

The measurement results for three prepared PDMS samples are presented. Figure 4.6 illustrates the force and indentation depth relation of the 1:30 homogeneous PDMS sample and the fitting curves based on Equation 4.2. Based on the fitting curves, the obtained elastic modulus of the sample at five discrete locations is listed in Table 4.3. The results achieved from the five

locations are very consistent; the standard deviation is only 4.3 kPa, smaller than 2% of the average value.

Table 4.3 Elastic modulus obtained from 1:30 homogeneous PDMS sample.

Transducers	1	2	3	4	5
Elastic modulus (Pa)	216,491	212,229	222,123	222,294	216,024



Figure 4.6 Force and indentation depth relation of the 1:30 homogeneous PDMS sample and the fitting curves.

The measurement results for the heterogeneous 1:30 and 1:40 PDMS samples are presented in Table 4.4. Figure 4.7 shows the curve fitting of the heterogeneous 1:40 PDMS measurement results. For the 1:30 heterogeneous sample, the third transducer reports the largest elastic modulus, which indicates that a heterogeneity with a higher elastic modulus exists above the third transducer. This result is expected, since the embedded corn crumble is in the middle area of the sample, as can be seen in Figure 4.3 (b). The rest of the transducers report values that are consistent with the homogeneous 1:30 PDMS sample, with an average modulus of 183.62 kPa. The value is slightly smaller than the one obtained from the homogeneous 1:30 PDMS sample, which is believed to be due to the fact that the solidification is affected by the embedded hard crumbles. The third row of Table 4.4 shows the characterization results of the heterogeneous 1:40 PDMS sample. Results from transducers 3 and 5 show the largest values, which indicates the heterogeneity with a higher elastic modulus exists at those locations. Figure 4.3 (c) shows the photo of the heterogeneous 1:40 sample being tested. From the photo, it can be seen that two large crumbles are located in the middle and along the right edge, which are believed to cause the increase in the elastic modulus at these two locations. A smaller crumble is at the left edge, which does not cause much change in the output. It is believed that this crumble is not embedded deep enough to cause a noticeable change in the device output, based on the simulation work done in Section 3.3.



Figure 4.7 Force and indentation depth relation of the 1:40 heterogeneous PDMS sample and the fitting curves.

	1st	2nd	3rd	4th	5th
1:30 heterogeneous (Pa)	185,682	166,514	252,861	175,485	206,796
1:40 heterogeneous (Pa)	84,348	79,610	102,985	80,877	104,963

Table 4.4 Elastic modulus obtained from the 1:30 and 1:40 heterogeneous PDMS samples through CSM.

Table 4.5 Measured elastic modulus of PDMS with 1:30 and 1:40 mixing ratios in the literature.

PDMS	[81]	[82]	[83]	[84]	[85]	[86]
1:30	N/A	0.42~1.53	304.6 kPa	~200 kPa	270 kPa	N/A
		MPa				
1:40	78 kPa	N/A	97.6 kPa	41.89 kPa	N/A	0.1 MPa

Table 4.5 presents the measurement results of 1:30 and 1:40 homogeneous PDMS samples from the literature review. The methods being used to achieve the results include nanoindentation [85], cyclic unconfined compression testing [83], and uniaxial mechanical characterization [84], etc. By comparing the PDMS elastic modulus measured by this experimental technique and the results obtained by other groups, it can be seen that the results are within the right range, which further indicates the reliability of the CSM technique for the elastic modulus measurement through a microfluidic device.

4.3.2.2 Measurement results for Agar samples

In this section, the measurement results for agar samples are presented. Unlike the PDMS samples, the agar samples exhibit a more obvious indentation depth-dependent property. A similar phenomenon was observed in biological tissues, as the modulus is dependent on the strain level [29]. In this case, instead of obtaining an overall elastic modulus through a regression method, the elastic modulus at each indentation depth is calculated. Since agar samples were tested in an aqueous condition, it took longer to achieve its steady state. Thus, to achieve a relatively stable

result, the data used in the data analysis are those after 5 seconds relaxation in each indentation.

Figure 4.8 shows the measured elastic modulus of the homogeneous 1%, 2%, and 3% agar samples at different indentation depths with error bars. Table 4.6 summarizes the results obtained from Figure 4.8.



Figure 4.8 Measured elastic modulus of the homogeneous (a) 1% agar, (b) 2% agar, and (c) 3% agar samples as a function of indentation depth.



As can be seen, the obtained elastic modulus for each sample is within a range. To better compare the differences between the three agar samples, the data corresponding to the same overall indentation depth 120 μ m to 300 μ m are used for comparison. Slight deviations among the five locations are observed from results of the 1% and 3% agar samples, which indicates the homogeneity of the agar samples. A larger variance exists in the measurement results of the 2% agar sample; this is believed to be caused by the misalignment of the probe, since an almost linearly decreasing trend is observed from the first to the fifth transducers. An average value within the range is also calculated for each sample, as can be seen in the last column of Table 4.6.

	1st	2nd	3rd	4th	5th	Average
1% agar (kPa)	26~36	26~37	24~34	25~32	24~32	30
2% agar (kPa)	86~139	65~121	58~98	58~99	57~94	87
3% agar (kPa)	88~171	87~165	93~171	100~176	108~175	144

Table 4.6 Elastic modulus obtained from the homogeneous agar samples with different concentrations through the CSM technique.

Table 4.7 Elastic modulus of the agar samples with different concentrations in the literature.

	[87]	[88]	[89]	[90]	[91]	[92]	[93]	[31]
1% (kPa)	5~6.6	N/A	15~24	N/A	N/A	N/A	13.2	~150
2% (kPa)	N/A	N/A	113~163	N/A	41.6	25.02 33.54	53.4	~375
3% (kPa)	100~132	52~499	N/A	122~155	N/A	80.81 86.07	257.9	N/A

Table 4.7 summarizes the characterization results of homogeneous agar gels in the literature, utilizing different measurement methods, including uniaxial quasi-static compression tests [87, 89], magnetic resonance elastography (MRE) [92], dynamic mechanical analysis (DMA) method [92], etc. From Table 4.7, it can be seen that the elastic modulus values of agar gels of different concentrations vary a lot among the various methods and groups. Results obtained by this work are comparable to the results obtained by others, except for the 1% agar sample, the result obtained is slightly larger than most of the other groups.

Figure 4.9 (a) to (c) presents the measurement results for 1%, 2%, and 3% heterogeneous agar samples. For the 2% heterogeneous agar sample, the results achieved from the third transducer show a much larger value than the rest, which indicates the existence of heterogeneity with a higher elastic modulus at that location. The picture of the 2% heterogeneous agar sample verified the conclusion. The small hard sphere is close to the middle of the sample, which is corresponding to the location of the 3rd transducer.



Figure 4.9 Measured elastic modulus of the heterogeneous (a) 1% agar, (b) 2% agar, and (c) 3% agar samples as a function of the indentation depth.



Figure 4.9 (a) also indicates higher elastic modulus exists close to the third transducer, since the elastic modulus obtained from the third transducer is larger than the rest, while the modulus extracted from other transducers' output is very close to each other.

From Figure 4.9 (c), it can be seen that the elastic modulus curve (green curve) calculated from the voltage output of the fourth transducer gives a larger elasticitic modulus, while the rest curves show a decrease in an orderly manner at the same indentation depth. This result indicates despite the misalignment of the indentation probe, which is concluded from the decreasing trend at each indentation depth, a larger elastic modulus exists at the location of the fourth transducer. Otherwise, the curve obtained from the fourth transducer should follow the trend, appearing between the red and dark blue lines.

4.4 Discussion and Conclusions

For simplicity, many assumptions are made for extracting the elastic modulus of the soft materials. First, it is assumed that the deformation of the testing materials is still in their linear regime, which means the strain level is within 5% to 10%. Secondly, it is assumed that the material follows elastic deformation rules. However, many other models are adopted to better describe the elastic property of soft materials, since most bio-materials start to have nonlinear behavior at higher strain rate. For example, hyperelastic material models have been used to model joint cartilage. Other models include neo-Hookean, Mooney-Rivlin, Saint-Venant Kirchhoff, Arruda-Boyce, Ogden, etc. These models could be considered for incorporation into Equation 4.2 in future work.

Overall, this chapter presents the implementation of the CSM technique on the elasticitic modulus measurement of both homogeneous and heterogeneous synthetic materials. The measurement results agree well with those presented in the literature. The capability of the device to measure the spatially-varying elastic modulus of soft materials through one measurement is successfully proved.

CHAPTER 5

CONCURRENT SPATIAL MAPPING OF THE LOSS MODULUS OF SOFT MATERIALS: THEORY

The viscoelastic behavior of a material is usually studied through the creep test, relaxation test and harmonic test. It has been successfully proven that the polymer-based microfluidic device is capable of detecting the spatially-varying viscous behavior of a material through a relaxation test [94]. This chapter and the following chapter explore the feasibility of measuring the viscous property of a material through decoupling its stiffness and damping coefficient from two sets of system-level parameters measured by the harmonic test, and estimate its spatially-varying loss modulus through the spatially-varying phase shifts. This chapter first presents the rationale related to the CSM of the viscoelastic behavior of soft materials via the polymer-based microfluidic device. Then, the derivation of relating the damping coefficient to the loss modulus of a material is presented. The experimental implementation will be provided in the next chapter.

5.1 Rationale

When a piece of material is under a sinusoidal loading, the response of the material, the phase shift in particular, can be used to detect the viscosity within a material. In Figure 5.1, the phase shifts between stress and strain in three different scenarios are illustrated. When the material is purely elastic, the stress and strain are completely in phase, while a viscous material gives an out-of-phase response. For viscoelastic materials, the phase lag is between 0 to 90°[95]. As such, phase shifts between the applied load and the corresponding deflection can manifest the viscous property of materials.



Figure 5.1 Illustration of phase shifts between stress and strain of (a) purely elastic material, (b) viscous material and (c) viscoelastic material under harmonic excitation.

Figure 5.2 schematically shows the rationale of applying the CSM technique to conduct a dynamic measurement. In a measurement, a rigid cylindrical probe is employed to exert a macroscopic sinusoidal load, $F=F_0 \cdot sin(\omega t)$, to a sample placed on the device. The probe displacement accompanying the load can be expressed as $z=z_0 \cdot sin(\omega t)$.

Due to the viscous component of the testing material, a phase shift will be generated between the input force and the sample deflection. The sample deflection can be expressed as:

$$z_{s-i} = z_{s0-i} \cdot \sin(\omega t + \delta_i) \tag{5.1}$$

where z_{s-i} denotes the deflection of a sample at the location of the *i*th transducer; δ_i is the phase difference caused by the viscous property of the sample.



Figure 5.2 Schematic of the rationale for measuring the heterogeneous viscous behavior of a piece of soft material.

Caused by the heterogeneity within a sample, at different locations, δ_i may have different values. When the deflection comes to the device, another phase shift θ_i is introduced. θ_i is caused by the PDMS structure of the device and the electrolyte in the microchannel. Ideally, θ_i should be the same at the five locations. However, due to the fabrication variation of the microchannel height at the five locations and the misalignment of the probe which is employed to apply the mechanical stimuli, θ_i will exhibit some variations among the five transducers. As a result, the final deflection detected by the *i*th transducer can be expressed as:

$$z_{s-i} = z_{s0-i} \cdot \sin(\omega t + \theta_i + \delta_i). \tag{5.2}$$

The phase shift detected by the device is a combination of θ_i and δ_i , as expressed in Equation 5.3.

$$\varphi_i = \theta_i + \delta_i \tag{5.3}$$

Since the variation of the phase shifts caused by the device itself is assumed to be negligible, the phase shift value φ_i eventually manifests the viscous property of the testing materials at the locations of the transducers. Thus, in this study, the phase shift in a sample along its length is represented by φ_i , and is used to estimate the viscous property distribution along the sample length.

In the experimental process, as will be described in detail in the next chapter, dynamic characterization of the device itself should first be carried out to extract the system-level parameters. The polymer-based microfluidic device under external excitation is modeled as a second-order mechanical system, as illustrated in Figure 5.3.



Figure 5.3 (a) Schematic of the side view of a microfluidic device under external sinusoidal loading, and (b) a second-order mechanical system model of the polymer-based microfluidic device under a sinusoidal excitation.
The dynamic behavior of the mechanical system is governed by the following equation:

$$M_D \cdot \frac{d^2 z}{dt^2} + C_D \cdot \frac{dz}{dt} + K_D \cdot z = F(t), \qquad (5.4)$$

where M_D , C_D , and K_D denote the equivalent mass, damping coefficient and stiffness of the system, respectively. *F* denotes the external excitation, which is the superposition of a harmonic load and an initial static load, as expressed in Equation 5.5.

$$F(t) = F_{initial} + F_0 \sin(\omega t)$$
(5.5)

where $F_{initial}$ is a compressive load that is initially imposed on the microfluidic device to ensure the device to undergo compressive load during the vibration cycle, and F_0 is the amplitude of the harmonic load. The corresponding deformation of the device can be expressed as:

$$z(t) = z_{initial} + z_0 \sin(\omega t + \varphi)$$
(5.6)

where $z_{initial}$ is the pre-deformation of the device under the initial load; z_0 denotes the amplitude of the harmonic deformation response of the device; and φ is the phase shift between the external excitation and the device response. φ is caused by the intrinsic damping of the device and other components in the system, and is investigated in a previous publication [96].

$$K_D - M_D \cdot \omega^2 = \frac{F_0}{z_0} \cdot \cos \varphi \tag{5.7}$$

$$C_D \cdot \omega = \frac{F_0}{z_0} \cdot \sin \varphi \tag{5.8}$$

Based on Equation 5.4, Equations 5.7 and 5.8 are derived, which relate the system- level parameters M_D , C_D , and K_D to the direct measurement values F_0 , z_0 and φ . Thus, the system-level parameters can be further calculated.

Next, a piece of viscoelastic material is inserted between the device and the probe, as illustrated in Figure 5.4 (a). As depicted in the dashed line frame in Figure 5.4 (b), the soft material and the device form a new mechanical system, with two springs and two dampers connecting in

series. C_S and K_S denote the damping coefficient and the stiffness of the sample, respectively. As can be seen in Figure 5.4 (b), the mechanical system in the dashed frame can be simplified as a spring with a combined stiffness K' and a damper with a combined damping coefficient C'connecting in parallel. Through the dynamic measurement, the combined stiffness K' and combined damping coefficient C' can be achieved by fitting Equation 5.9 through the nonlinear regression method.



Figure 5.4 (a) Schematic of the side view of a piece of soft material under the dynamic measurement, and (b) a second-order mechanical system model of a piece of soft material under a dynamic measurement.

$$\frac{z_0}{F_0} = \frac{1}{\sqrt{(K' - M' \cdot \omega^2)^2 + C'^2 \cdot \omega^2}}$$
(5.9)

To decouple the material properties of the test material from the combined stiffness and damping coefficient, Equations 5.10 and 5.11 are applied.

$$K_{S} = \frac{1}{\frac{1}{K'} - \frac{1}{K_{D}}}$$
(5.10)

$$C_{S} = \frac{1}{\frac{1}{C'} - \frac{1}{C_{D}}}$$
(5.11)

5.2 Relation between the Loss Modulus and the Damping Coefficient

The energy loss in a material sample within one harmonic vibration cycle can be expressed as:

$$W_d = \oint F_d dx = \oint C\dot{x}dx = \oint C\dot{x}^2 dt \qquad (5.12)$$

where F_d denotes the damping force; *C* denotes the damping coefficient of the material; and *x* denotes the harmonic displacement imposed on the sample.

The imposed displacement on the sample can be expressed as:

$$x = X\sin(\omega t - \delta) \tag{5.13}$$

where *X* denotes the displacement amplitude and φ is the phase between the displacement and the harmonic force input.

The vibration velocity applied to the sample can be expressed as the first derivative of the imposed displacement, as expressed in Equation 5.14.

$$\dot{x} = \omega X \cos(\omega t - \delta) \tag{5.14}$$

Thus, the energy dissipation during one vibration cycle can be further written into:

$$W_{d} = \oint F_{d} dx = \oint C \dot{x} dx = \oint C \dot{x}^{2} dt = C \omega^{2} X^{2} \int_{0}^{\frac{2\pi}{\omega}} \cos^{2}(\omega t - \delta) dt = \pi C \omega X^{2}$$
(5.15)

Assuming the cyclic force is uniformly applied to a material sample in a bulk shape, the stress and strain distributions are considered to be uniform throughout the entire material. Then the energy loss can be expressed as:

$$W_{d} = \iiint \int_{t_{1}}^{t_{1}+T} \sigma d\varepsilon dV = \oiint \int_{t}^{t+T} \sigma \frac{d\varepsilon}{dt} dt dV \qquad (5.16)$$

The stress, strain, and their relation can be further written into Equations 5.17 to 5.19:

$$\sigma = \sigma_0 e^{i\omega t} = \frac{F_0 e^{i\omega t}}{A} \tag{5.17}$$

$$\varepsilon = \varepsilon_0 e^{i(\omega t - \delta)} = \frac{X}{h} e^{i(\omega t - \delta)}$$
(5.18)

$$\sigma = E\varepsilon = (E' + iE'')\varepsilon \tag{5.19}$$

where A and h denote the cross section area and the thickness of the material sample, respectively; *E* is the complex modulus.

The complex modulus can be further written into:

$$E = \frac{\sigma}{\varepsilon} = \frac{\sigma_0}{\varepsilon_0} e^{i\delta} = E' + iE'' = \frac{F_0 h}{AX} \cos \delta + i \frac{F_0 h}{AX} \sin \delta$$
(5.20)

The real part of the strain and stress can be expressed as:

$$\varepsilon_R = \varepsilon_0 \cos(\omega t - \delta) \tag{5.21}$$

$$\sigma_R = \sigma_0 \cos(\omega t) \tag{5.22}$$

Then, the energy loss per unit volume during a vibration cycle can be expressed as:

$$w_{d} = \int_{t_{1}}^{t_{1}+T} \sigma_{R} d\varepsilon_{R} = \int_{t_{1}}^{t_{1}+T} \sigma_{R} \frac{d\varepsilon_{R}}{dt} dt$$

$$= -\omega \int_{t_{1}}^{t_{1}+T} \frac{F_{0}}{A} \cos(\omega t) \frac{X}{h} \sin(\omega t - \delta) dt$$

$$= -\omega \frac{F_{0}X}{Ah} \int_{t_{1}}^{t_{1}+T} \cos(\omega t) \sin(\omega t - \delta) dt$$

$$= -\omega \frac{F_{0}X}{Ah} \left(-\frac{\pi}{\omega}\right) \sin \delta = \frac{F_{0}X\pi\sin\delta}{Ah}$$

(5.23)

For the entire sample, the energy loss per cycle can be expressed as:

$$W_d = \iiint w_d dV = F_0 X \pi \sin \delta = \frac{F_0 h}{AX} \sin \delta \cdot \frac{X^2 A \pi}{h} = E'' \frac{X^2}{h} A \pi = \pi C \omega X^2 \quad (5.24)$$

From here, the loss modulus can be related to the damping coefficient as:

$$E'' = \frac{C\omega h}{A} \tag{5.25}$$

Equation 5.25 calculates the overall loss modulus of a material. In conjunction with the obtained phase shifts along the sample length, the distribution of the loss modulus along the sample length can be estimated.

According to Equation 5.20, the relation between the storage modulus and loss modulus can be expressed as

$$\frac{E''}{E'} = \tan \delta \,. \tag{5.26}$$

Thus the storage modulus can be written into:

$$E' = \frac{E''}{\tan \delta}.$$
 (5.27)

CHAPTER 6

CONCURRENT SPATIAL MAPPING OF THE LOSS MODULUS OF SOFT MATERIALS: IMPLEMENTATION

In this chapter, the implementation of the CSM technique for measuring the spatiallyvarying loss modulus of soft materials via the microfluidic device is presented. The experimental setup, protocol, data analysis methods, and measurement results are presented in detail.

6.1 Materials and Methods

6.1.1 Experimental setup

The experimental setup is shown in Figure 6.1, which mainly includes a steel frame, a shaker assembled to the frame, a load cell, a distance sensor, and a microfluidic device mounted on a 5-axis manipulator.



Figure 6.1 Experimental setup.

The laser distance sensor CP08MHT80 is also fixed to the frame and is placed upside down to avoid interactions from other light sources. It is used to measure the vibration amplitude from the shaker. A cone-shaped structure is placed in the setup for reflection of the laser beam emitted from the distance sensor. A structure with a cone shape is chosen, because of its high stiffness level compared to a circular plate, or a cantilever, in order to minimize the flexural mode vibration under the excitation along the vertical direction.

The voltage output of the distance sensor should be within the range of 0-10 V, and the resolution of it is less than 8 μ m. The load cell assembled to the setup is ATI nano 17, the smallest commercially available 6-axis transducer, with a 1/320 N resolution along the z direction.

Compared to the device employed in the CSM for elastic modulus measurement presented in Chapter 4, the microfluidic device employed in this setup has a smaller transducer spacing and thicker PDMS structure, as shown in Figure 6.2. It is referred to as device #2 in Chapter Two, whose dimensions are listed in the third column of Table 2.1.



Figure 6.2 A microfluidic device used in the loss modulus measurement of soft materials through CSM.

Figure 6.3 schematically illustrates the experimental setup and signal flow for conducting the loss modulus measurement through CSM. The vibration shaker is controlled by a sinusoidal signal generated in a data acquisition (DAQ) board through LabVIEW, with 25000 samples per second. The distance sensor mounted to the frame measures the vibration amplitude of the cone-shaped structure, which is connected to the shaker through a rigid connection. The voltage output of the device is sent to the DAQ board and analyzed by a LabVIEW program which averages the vibration amplitude within five seconds and sends it to a proportional-integral-derivative controller (PID) module. The PID module adjusts the signal amplitude for the shaker control accordingly to make sure the amplitude of the vibration generated by the shaker remains at a desired constant value among all the frequencies. The direct output from the device.



Figure 6.3 Schematic of the experimental setup and signal flow for the loss modulus measurement of soft materials through CSM.

The load cell is connected between the vibration shaker and the cone-shaped laser reflector which also accounts for the majority of the equivalent mass of the vibration system. The device is placed on the five-axis manipulator which helps adjust the orientation of the device to achieve better alignment. A block with a cylinder probe is connected to the laser reflector, which is used to exert the amplitude-demodulated vibration to a sample placed on the device. The output from the load cell and the device is recorded by the DAQ board through another LabVIEW program. The sampling rate is set to be 5000 for all the dynamic measurement with this setup.

6.1.2 Sample preparation

In the measurement, materials obtained from Smooth-on Inc. are used as the homogeneous material samples. PDMS with a 1:30 agent to base mixing ratio is used as the matrix material for making heterogeneous samples. Similarly, heterogeneity is created by embedding corn crumbles into the PDMS samples. The dimensions of the samples are shown in Table 6.1.

The photos of the testing samples are shown in Figure 6.4. As can be seen, the samples were cut into rectangular prisms, yet the dimensions vary from sample to sample.

Sample name	Dimensions (mm)
	$l \times w \times h$
Dragon skin pro	8.13×5.59×3.56
Dragon skin 10	9.40×5.10×2.54
Dragon skin 20	9.10×5.10×3.81
Ecoflex	8.89×5.33×3.30
1:30 PDMS 1	12.95×4.83×3.05
1:30 PDMS 2	8.38×4.06×3.05

Table 6.1 Dimensions of the testing samples.





Figure 6.4 Photos of the samples: (a) dragon skin 10, (b) dragon skin 20, (c) Ecoflex, (d) dragon skin FX-Pro, (e) the first 1:30 PDMS sample with heterogeneity inside, and (f) the second 1:30 PDMS sample with heterogeneity inside.

6.1.3 Experimental protocol

Similar to using the CSM technique to measure the spatially-varying elastic modulus in Chapter 4, the experimental protocol for using the CSM to measure the loss modulus of soft materials also includes three steps: probe alignment, control experiment, and dynamic CSM of materials.

a) Probe alignment

Probe alignment should first be carried out to minimize errors caused by misalignment of the probe in the output. This process is completed with the help of a five-axis manipulator and a LabVIEW program that displays the output. The probe which is connected with the shaker is first visually placed above the device, parallel to the microchannel. Since the position of the probe is fixed, the position of the device is adjusted by the five-axis manipulator to reach a better alignment. For observing the alignment, the device is moved up by the five-axis manipulator when it is approximately 0.5 mm away from the probe until it is in touch with the probe. At the same time, the output signals from the five transducers of the device are recorded by the LabVIEW program and exhibited on the screen in the LabVIEW front panel. Through the five-axis manipulator, the orientation of the device is further adjusted to make sure the change of the five voltage outputs happens at the same time, which indicates a proper alignment. After the orientation is fixed, the device is brought down, so that no initial deformation is applied to the device at this stage.

Although the misalignment of a probe is alleviated by using voltage change in the calculation for obtaining the elastic property of a material, as mentioned in Section 2.4 and Chapter 4, it has a larger effect on the phase shift.

Evidently, a better alignment will improve the accuracy and reproducibility of the measured results. Although great care has been taken to conduct the probe alignment, misalignment still more or less exists.

b) Control experiment

A control experiment is conducted on the device itself prior to measuring a sample, in order to extract the system-level parameters for future use. Since the device itself is a damped mechanical system, it is expected to observe a phase shift between the load input and the voltage output of the device.

After the alignment is finished, the device is brought up towards the probe to generate a uniform pre-deformation, based on which a sinusoidal load is applied by the shaker. The frequency varies from 5 to 50 Hz, with an interval of 1 Hz. The device is pre-deformed by 300 μ m, to ensure the characterization is based on compressive oscillation. With the help of a feedback from the distance sensor to the PID module, the vibration amplitude generated by the shaker is fixed at 100 μ m among all the frequencies. The sampling rate for data acquisition is set to be 5000.

Right after the control experiment, the prepared samples are inserted between the device and the probe one at a time. In the measurement, a pre-deformation between 300 to 700 μ m is imposed on the samples, to ensure a clear and complete signal output from the device. For consistency, the vibration amplitude is fixed at 100 μ m for testing all the samples. The frequency range is 5 to 40 Hz, and the frequency interval is 1 Hz. The sampling rate for data acquisition is also set to be 5000.

6.2 Data Analysis and Results

6.2.1 Data analysis

Waveforms can be represented by a series of harmonic functions with different frequencies, among which Fourier series is one of the representations [97]. Equation 5.4 is the expression of the Fourier series.

$$\mathbf{x}(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos(2\pi n f_1 t) + b_n \sin(2\pi n f_1 t)$$
 (6.1)

Fourier transform is commonly used in engineering to transform signals from time domain to frequency domain and the discrete Fourier transform (DFT) allows the discrete experimental data to be fitted into Fourier series to obtain the fundamental frequency components. The Fast Fourier Transform (FFT) is an algorithm to compute the DFT; most numerical methods in frequency domain are based on this algorithm. Equation 6.2 illustrates the transform from the time domain to the frequency domain, and Equation 6.3 represents this transform when discrete data are used.

$$F(\omega) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} f(t) e^{-i\omega t} dt \qquad (6.2)$$

$$X_{k} = \sum_{n=0}^{N-1} x_{n} e^{-i2nk\frac{n}{N}}$$
(6.3)

Due to the viscous component in the samples, there is a phase shift δ between the harmonic force and the corresponding deflection of the samples. It can be seen from the data in the time domain, the voltage output from the five transducers which represents the deformation of the microchannel is behind the voltage output from the load cell.

In this section, a numerical method based on fast Fourier transform (FFT) is used to obtain the phase shift between the device output and the applied force using MATLAB®. Assuming s_1 and s_2 are the data series after applying fast Fourier transform, the expression for obtaining the cross spectrum is shown in Equation 6.4.

$$Cross spectrum = s_1 \times conj(s_2) \tag{6.4}$$

In this study, phase shifts are calculated by carrying out cross spectrum analysis.

In Equations 5.7 and 5.8, M_D , K_D , and C_D are unknowns, while F_0 , z_0 and φ can be obtained from the experimental data. Equation 5.7 shows a linear relation between the dynamic stiffness and ω^2 . Through a linear curve fitting, the unknown parameters M_D and K_D can be obtained. Similarly, a linear relation also exists between the dynamic damping and the angular frequency ω . The damping coefficient C_D of the system also can be obtained using a linear curve fitting.

To achieve the amplitude of the external load F_0 and the vibration amplitude z_0 , the amplitudes at each cycle are calculated and an average is taken among all the cycles in the data. For extracting the system-level parameters as a combination of the device and a material, a nonlinear regression method is applied to calculate M', K' and C' from Equation 5.9.

6.2.2 Control experiment results

Figure 6.5 plots the system response obtained from the dynamic characterization of the system with the microfluidic device. From Figure 6.5, it can be seen that resonance happens around 34 Hz.



Figure 6.5 Phase and the amplitude ratio obtained from the dynamic characterization of the system with the setup and the microfluidic device.

Figure 6.6 and Figure 6.7 are the curve fitting results of the system with the device and the setup. Table 6.2 exhibits the calculated values of the system-level parameters of the system with the microfluidic device, based on the linear curve fittings.



Figure 6.6 Dynamic stiffness of the system as a function of ω^2 .



Figure 6.7 Dynamic damping of the system as a function of ω .

 K_D (N/m)12637 M_D (kg)0.2096 C_D (Ns/m)36.457

Table 6.2 System-level parameters of the dynamic system with the microfluidic device and the setup, based on the linear curve fitting.

Based on these values, the natural frequency, damping ratio, damped frequency, and the quality factor of the system are calculated, and these values are listed in Table 6.3.

System level	values
properties	
f_0 (Hz)	39.04
ζ	0.354
f_d (Hz)	34.14
Q	1.41

Table 6.3 System-level properties of the dynamic system.

6.2.3 Measurement results of the samples

After extracting the system level parameters of the dynamic system with the device alone, the measurement results with a piece of test sample in the dynamic system are also achieved. Figure 6.8 plots the measurement results of the Dragon Skin 10 sample. As can be seen from both the phase and compliance diagrams, the resonance happens at around 23 Hz. Compared with the damped resonance value obtained previously in the control experiment without the sample, which is around 34 Hz, the resonance frequency decreased. The decrease is caused by the change of contact stiffness, which further contains information of the mechanical property of the unknown sample. In Figure 6.8 (a), the phase shift obtained from the experimental data at around 40 Hz exceeds 180 °. It is possibly caused by a time delay during the sequential data acquisition process.



Figure 6.8 Dynamic measurement results of (a) phase and frequency relation, and (b) compliance and frequency relation of the dynamic system with Dragon Skin 10.

Equation 5.9 is fitted using a nonlinear regression method to obtain the equivalent stiffness, damping coefficient, and mass of the new dynamic system with the sample. With the new contact, the equivalent stiffness, damping coefficient of all samples are calculated and listed in Table 6.4.

The change of the equivalent mass caused by the new contact is ignored, since the majority of the mass is caused by the cone-shaped structure.

Based on Equations 5.10 and 5.11, the stiffness and the damping coefficient of the sample itself is decoupled. The decoupled results are presented in Table 6.5. From the last column of the table, it is noticed that the thickness of the samples varies. This is because during the sample preparation, the thickness of the sample is not strictly controlled. To account for this issue, the results are normalized, assuming the thickness is 3 mm for all the materials. Equation 6.5 relates the stiffness to the elastic modulus of a material with its dimensions, under uniaxial test. In Equation 6.5, E is the elastic modulus of the material, A is the area of the sample cross section, and h is the sample thickness. Since E is a constant, k is inversely proportional to the thickness of a sample. With the help of Equation 6.5, the decoupled stiffness is normalized and listed in Table 6.6.

$$k = E\frac{A}{h} \tag{6.5}$$

	Combined equivalent Stiffness	Combined equivalent damping coefficient C'(N, a/m)	Thickness (mm)
David and all the same	<u>A (IN/III)</u>	<u> </u>	2.50
Dragon skin pro	994.26	5.623	3.56
Dragon skin 10	1994.3	4.399	2.54
Dragon skin 20	1560.6	5.154	3.81
Ecoflex	941.24	6.982	3.3
1:30 PDMS	3685.9	6.018	3.05
1:30 PDMS 2	1608.8	6.275	3.05

Table 6.4 Measurement results of the dynamic system with a piece of testing material.

A similar relation exists between the damping coefficient of a sample and its thickness, as can be seen in Equation 6.6. The derivation is shown in Section 5.2. The damping coefficient is normalized to 3 mm for all the samples based on Equation 6.6, which is also shown in Table 6.6.

$$C = \frac{E''A}{\omega h} \tag{6.6}$$

Decoupled Stiffness k_s	Decoupled damping	Thickness (mm)
(11/111)	$(N \cdot s/m)$	
1079.2	6.648	3.56
2368.0	5.003	2.54
1780.5	6.003	3.81
1017.0	8.636	3.3
5203.7	7.208	3.05
1843.5	7.580	3.05
	Decoupled Stiffness <i>k</i> s (N/m) 1079.2 2368.0 1780.5 1017.0 5203.7 1843.5	Decoupled Decoupled Stiffness k_s damping (N/m) coefficient c_s (N·s/m) (N·s/m) 1079.2 6.648 2368.0 5.003 1780.5 6.003 1017.0 8.636 5203.7 7.208 1843.5 7.580

Table 6.5 Decoupled results of the testing materials.

The Shore hardness and the mixed viscosity of the materials provided by Smooth-on Company are also provided in the last two columns of Table 6.6. As can be seen from the second and the fourth columns, the stiffness of the materials corresponds well with their Shore hardness, since hardness and stiffness are both reflections of the elastic modulus of materials. The measured stiffness of the prepared PDMS samples indicates the sample with more hard components (corn crumbles) embedded is much stiffer than the other PDMS sample (1:30 PDMS 2) which only has one corn crumble embedded. Interestingly, the damping coefficient of the polymer materials does not follow the trend of the mixed viscosity. Ecoflex is the softest material according to the experimental results, and it has the largest damping coefficient. However, the mixed viscosity within the material is the smallest, which indicates the viscoelastic damping of a material is not just proportional to the viscous ingredient being mixed with the other components to make the polymer. The comparison between the two PDMS samples successfully proves the rigid components inside a sample can reduce the overall viscous damping of a material.

	Normalized Stiffness k _s ' (N/m)	Normalized damping coefficient c_s ' (N·s/m)	Shore hardness	Mixed viscosity (cps)
Dragon skin pro	1280.6	6.67	2A	18000
Dragon skin 10	2004.9	3.72	10A	23000
Dragon skin 20	2261.2	6.55	20A	20000
Ecoflex	1118.7	7.68	10	14000
1:30 PDMS 1	5290.4	6.12	-	-
1:30 PDMS 2	1874.2	6.38	-	-

Table 6.6 Measurement results of the samples after normalization with comparison to the information provided by Smooth-on Inc.

Equation 6.6 also relates the loss modulus with the damping coefficient C, as it can be rewritten into Equation 6.7.

$$E'' = \frac{Ch\omega}{A} \tag{6.7}$$

In this study, it is assumed the damping coefficient *C* is a constant among different frequencies. Thus, the loss modulus *E* " linearly increases with the angular frequency, as plotted in Figure 6.9. As can be seen in the plot, when the frequency is lower than 10 Hz, the loss modulus of most of the test materials are below 20 kPa. With the increase of the frequency, the loss modulus of the materials increases as well, which demonstrates that the intrinsic viscous property of a material increases the damping when the loading frequency increases. The increasing trend is comparable to the results achieved in the work done by Placet [98] for the dynamic characterization of PDMS material. It is reported that PDMS with 10 to 1 polymer to agent ratio has the loss modulus of 0.02 to 1.6 MPa, at the frequency range 0.01 to 105 Hz.



Figure 6.9 Calculated loss modulus of the testing materials.

As mentioned in the previous chapter, phase shifts between the overall load and the voltage output from the five transducers can manifest the distribution of the viscous property of a material along the sample length. Figure 6.10 plots the relation between the phase shift and frequency of the two heterogeneous 1:30 PDMS samples. Table 6.7 displays the phase shifts of the two heterogeneous PDMS samples at the frequency of 18 Hz. The phase shift from each transducer location is normalized with the average value of the five phases. The distribution of the loss modulus of the two samples at the frequency of 18 Hz is estimated and listed in Table 6.8, by multiplying the normalization coefficients.



Figure 6.10 Phase shift between the force and the voltage output from the device with a 0.75 mm transducer spacing, when (a) heterogeneous 1:30 PDMS sample1, (b) heterogeneous 1:30 PDMS sample2 are tested.

	1	2	3	4	5
PDMS 1 (°)	32.82	35.23	38.80	41.10	44.41
Normalization	0.853	0.916	1.009	1.068	1.154
coefficient					
PDMS 2 (°)	44.52	44.27	44.61	45.40	49.81
Normalization	0.974	0.968	0.976	0.993	1.090
coefficient					

Table 6.7 Phase shifts of the two 1:30 PDMS heterogeneous samples at the frequency of 18 Hz and their normalized values.

Table 6.8 Estimation of the loss modulus distribution of the two 1:30 PDMS heterogeneous samples at the frequency of 18 Hz.

	1	2	3	4	5	
PDMS 1 (kPa)	28.78	30.91	34.04	36.03	38.94	
PDMS 2 (kPa)	63.00	62.61	63.13	64.23	70.50	

So far, the feasibility of using the CSM technique through a microfluidic device for the loss modulus measurement is proven. It is also noticed that the phase shifts measured by this device is not as obvious as the results measured with the device with a larger transducer spacing. It is believed to be caused by the thick PDMS structure and the small transducer spacing. However, it was the only device that was available when the measurement was conducted. Since the feasibility of the CSM technique for the measurement of the spatially-varying loss modulus is already proven, no further attempt was made to fabricate a new device. In Appendix E, the spatially-varying phase shifts measurement conducted through a device with a larger transducer spacing is presented. Back then, no distance sensor was available. Thus, only the phase shifts were extracted from the dynamic measurement. The phase shifts clearly shows the heterogeneity location within a sample.

From Equation 5.20, the relation between the storage modulus and the loss modulus can be related as:

$$\frac{E''}{E'} = \tan\delta \tag{6.8}$$

where δ is the phase shift between the force and deflection of the material. However, the measured phase angle in this work is a combination of the device and the material $\varphi_i = \theta_i + \delta_i$, as mentioned in the previous chapter. Although the measured φ_i in this work is used to estimate the loss modulus distribution, the measurement of δ_i should be further explored to obtain an accurate estimation of the storage modulus distribution.

6.3 Discussion and Conclusions

Based on the CSM technique, both the stiffness and the damping coefficient of polymer materials are quantitatively calculated, and the loss modulus distribution along the heterogeneous PDMS sample length is estimated. Some thoughts and a discussion are given here, regarding the experimental technique, theoretical model, and improvement in the future.

1) The resonance of the dynamic system is dependent on the entire setup, not the device alone. This conclusion is achieved by comparing the results obtained in Section 0 and those in previous publications [96]. The system-level parameters of the device in the publication with a different setup are listed in Table 6.9. The resonance and damping coefficient vary drastically with different setups, yet the stiffness of the device does not differ much in the two setups. Therefore, the measurement should keep the setup consistent with both the control experiment and the test for all the samples.

Results achieved by previous setup	K_D (N/m)	M_D (kg)	C_D (Ns/m)
1st	3064	0.0060	2.54
2nd	2854	0.0056	2.42
3rd	3026	0.0059	2.50
4th	2627	0.0050	2.15
5th	2716	0.0049	2.21

Table 6.9 System level parameters of the dynamic system based on the linear curve fitting when a different setup is used.

- 2) The device and the probe forms the original contact, when a piece of soft material is added, a new contact between the probe and the material is formed. The new contact alters the resonance, as can be seen in the previous section. Furthermore, different materials exhibit different effects on altering the resonance. As expected, the test material reduced the entire stiffness of the system comprising of the shaker, probe, cone-shaped structure, and the device since the testing material is connected in series with the rest of the setup. As a result, the resonance of the system is reduced, and the damping coefficient also varies, from where the material property of the soft material can be extracted.
- 3) Figure 6.11 plots the frequency response of the free vibration of the setup with only the shaker, cone-shaped structure, and the probe, without any device or material in touch. The vibration amplitudes are 100 μm, 200 μm, and 300 μm, respectively. The averaged equivalent stiffness, mass, and damping coefficient are calculated as 375.7 N/m, 93.6 g, and 10.76 Ns/m. In this work, the effect of the setup itself is not taken into account in interpreting the experimental results for obtaining the material property. Although in the dynamic mechanical analysis (DMA) technique, the effect of the instrument itself is not taken into account in taken into consideration when interpreting the experimental data either [95], the

nanoindentation technique considers and removes the effect caused by the instrument in the data analysis [99, 100]. The experimental technique avoids bringing complexity by drawing attention to the comparison of two different contacts, it is believed both contain the effect of the instrument itself. The stiffness and damping effect caused by the instrument are believed to be eliminated by the decoupling calculation method.



Figure 6.11 Frequency response of the free vibration of the shaker and probe.

- 4) The material property is obtained by decoupling from the results measured based on two different contacts, which are the contact between the device and probe and the contact between the testing material and the probe. Different from testing the material alone on a rigid substrate with the setup, the effect caused by either the rigid substrate or the rest of the setup can be eliminated by using this calculation method.
- 5) The resonance along the vertical direction is close to 25 Hz when a piece of soft material

is under testing with the current setup. This value is lower than it is expected, since the frequency of the mechanical noise in the environment is around the value. To avoid this situation, the setup should be designed more rigidly and with less weight to increase its resonance in the future.

- 6) The formula used to achieve the loss modulus and to normalize the results is based on the assumption of uniaxial uniform loading. The actual loading condition differs from the assumption, which requires a more accurate theoretical model to relate the relations.
- 7) Figure 6.6 shows a good linear relation between the dynamic stiffness and the square of angular frequency, however Figure 6.7 does not show a good linear fitting. It can be attributed to fact that K and M of the system are constants, however, C actually varies among different frequencies. Instead of assuming C is a constant in this work, a function of frequency should be used to express C in the future.

Overall, along with the previous chapter, this chapter presents the implementation of the CSM technique for the measurement of the spatially-varying loss modulus of soft materials. This is the first attempt so far to measure this property using one single device through the harmonic test. The obtained stiffness of the homogeneous samples matches well with the hardness provided by the manufacturer. The obtained loss modulus of PDMS samples is within the range provided by other researchers.

CHAPTER 7

CONCLUSIONS AND FUTURE WORK

In this chapter, the highlights and conclusive remarks of this dissertation are presented, and recommendations for the future work are proposed based on the current achievement.

7.1 Discussion

There are several drawbacks associated with the current device and the experimental technique.

- 1. Cross talk between sensing units exists in the current microfluidic device.
- Several assumptions are used in the formula deduction of this work, such as the small strain assumption. Ideally, the strain should be smaller than 10% under this assumption. However, the actual strain applied may exceed the upper limit.
- 3. For now, the *in vivo* test still hasn't been achieved. Soft materials were cut into a thin slice, and put on top of the device for further measurement. As a result, for biological tissues, the measured material properties may be different from tissues in a living state.
- 4. The current experimental technique still relies on a controlled displacement applied to the soft material sample through an external instrument, which limits the application inside the lab.
- 5. The voltage drift issue is investigated but still remains unsolvable.

7.2 Conclusions

This work focuses on the application of a microfluidic device for distributed-load detection. The main achievement of this work includes the following aspects.

- 1. A detailed input and output relation of one sensing segment of the microfluidic device is presented. The controlled input to the device is usually an indentation depth applied by a probe. The relationship between the displacement and the output voltage of the device is studied both theoretically and numerically. The linear relation between the prescribed indentation depth and the reaction force and the linear relation between the prescribed indentation depth and the output voltage from the numerical analysis is consistent with the results obtained from the characterization experiment. Based on these relations, device design and optimization can be carried out.
- 2. For the first time, the technical issues associated with the device and the experimental technique, such as voltage drift and probe misalignment issues are thoroughly studied. The voltage drift issue is experimentally explored, and the misalignment issue is explained in detail using numerical simulation.
- 3. A mathematical model to relate the force and indentation depth of a rigid cylindrical probe pressing upon a flat piece of material with finite thickness is proposed. Thus, the foundation of using the microfluidic device to measure the elastic modulus of a flat piece of soft material through a cylindrical probe and a load cell is established.
- 4. Most importantly, this work proved the capability of using the microfluidic device to distinguish heterogeneity within a piece of soft material and quantitatively obtain its spatially-varying viscoelastic properties along the sample length from the indentation and dynamic tests, respectively.

The main contribution of this work is that it lays down the foundation for applying a microfluidic device to achieve the CSM of the viscoelastic properties of soft materials. This work shows the simplicity in the device fabrication, and the efficiency in the measurement. The samples

used in this work are polymers, including customized PDMS samples with heterogeneity and standard commercialized polymer samples that are widely used to mimic biological samples or build polymer-based sensors.

7.3 Future Work

1. Number of transducers

The devices employed in this work only involves five sensing segments for simplicity. Devices with sensing segment arrays should be explored to increase the number of transducers, so that it can measure the viscoelastic properties of soft materials at more locations within one measurement.

2. Crosstalk or coupling should be experimentally explored

In this work, weak crosstalk or coupling from both mechanical and electrical aspects is assumed. It is necessary to explore the crosstalk or coupling in the future for a better calibration of the device.

3. Improvement of the experimental setup

In the experimental setup exhibited in Chapter 6, the cone-shaped structure is connected below the load cell, which brings complexity to the vibration system. In the future work, structure with less weight should be designed and incorporated in the vibration system for reflecting the laser beam from the distance sensor.

4. Device design and optimization

In Chapter 2, the input and output relation of the device is established numerically. Based on the simulation, device design and optimization through building a surrogate model and applying optimization algorithm becomes achievable. The measurement range of the device should be quantified. Improvement should be made to increase the device sensitivity and the output linearity.

5. Flexible substrate

The current design of the microfluidic device has a rigid substrate, which helps maintain its mechanical stability, yet this limits its application, especially when PDMS is a flexible, biocompatible material. A device with a flexible substrate should be explored for obtaining a bendable device, which can be further attached to surfaces with curvature.

6. Better wire connections and packaging

The current device adopted a wire bonding technique from MEMS technology to make the electrical connections to a PCB board which further connects to the circuits. The bonding wire is thin and vulnerable; a small mechanical impact or accidental touch would cause a disconnection, which is inconvenient even for an *in vitro* test, and makes an *in vivo* test even more difficult. To solve this problem, a better solution for connections between the device and the rest of the circuit is very necessary.

7. Apply the methodology to biological tissues

The data collected in this work are based on surrogates that are commonly used to mimic biological tissues, such as PDMS, dragon skin, and agarose gels. The mechanical property of biological tissues, such as cartilage and blood vessels is of utmost research interest and the initial design interest of the microfluidic device. Since the capability of the CSM of the viscoelastic properties of soft materials is already proven by this work, future work should focus on applying this technique to biological samples and relating the obtained results to pathological observations. Improve the load/displacement actuation mechanism to achieve integrated design. As mentioned in Section 7.1, the external mechanical excitation is added

by lab instruments such as a micromanipulator or a vibration shaker in this experimental technique. In the future, actuation mechanisms should be integrated with the microfluidic device to achieve portable/wearable applications. Figure 7.1 depicts the conceptual configurations of force sensing elements being integrated to holders [44, 101]. Similarly, the microfluidic device being used in this work should also be integrated into a small holder with a proper actuation mechanism that can provide mechanical excitation or displacement control. Then, the *in vivo* CSM of the mechanical property of biological tissues will ultimately be achieved.



Figure 7.1 Conceptual drawings of (a) a portable force sensor with distributed force sensing element [44] and (b) conventional forceps mounted with a soft tactile sensor array [101].

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APPENDIX A

CALIBRATION OF WENGLOR® CP08MHT80 DISTANCE SENSOR

The distance sensor Wenglor® CP08MHT80 was first calibrated with a MP-285 Motorized Micromanipulator, prior to being incorporated into the experimental setup utilized for the dynamic measurement in Chapter 6. The micromanipulator has a moving part that can be precisely controlled by a LabVIEW program to move a certain distance each time.

The micromanipulator was programmed to move down by 20 µm at a speed of 1 mm/s and stay for 5 seconds each time. This motion was repeated 10 times. The voltage output from the distance sensor that represents the distance between the moving part of the micromanipulator and the laser source of the distance sensor was recorded. The blue line in Figure A.1 represents the original signal from the distance sensor, while the red line in Figure A.1 represents the voltage data after being filtered in MATLAB®. An average was taken when the probe of the micromanipulator was at a steady state for each movement, based on the filtered data. The averaged voltage values are listed in Table A.1.

Time period (sec)	Voltage output (V)	Time period (sec)	Voltage output (V)
0~5	2.4247	25~30	2.4022
5~10	2.4190	30~35	2.3984
10~15	2.4133	35~40	2.3948
15~20	2.4094	40~45	2.3910
20~25	2.4056	45~50	2.3874

Table A.1 Averaged voltage values as the micromanipulator staying still at each motion cycle.



Figure A.1 Voltage output from the disptance sensor.

Assuming the micromanipulator's motion strictly follows the desired pattern, which is moving down 20 µm and staying still for 5 seconds, then the difference in distance between two time periods should always be 20 µm. Then, we can assume $\Delta x = x_i - x_{i-1} = 20$ µm in Equation A.1, and obtain a coefficient between each two time periods that relates the voltage change and the change in displacement, as can be seen in Table A.2.

$$c_i = \frac{V_i - V_{i-1}}{x_i - x_{i-1}}$$
(A.1)

The nominal coefficient that is used to relate the voltage change and displacement change is 0.2 V/mm. The averaged value for coefficient c_i is 0.207 V/mm, based on the characterization of the distance sensor. The error can be attributed to be from the micromanipulator itself, since 20

Time period (sec)	c_i (V/mm)	Time period (sec)	c_i (V/mm)
0~5	-	25~30	0.168
5~10	0.285	30~35	0.194
10~15	0.284	35~40	0.179
15~20	0.192	40~45	0.187
20~25	0.192	45~50	0.179

Table A.2 Ratio between voltage change and displacement change between two neighboring cycles.

APPENDIX B

HEAT EQUILIBRIUM TEST FOR THE MICROFLUIDIC DEVICE

In the test, the third transducer of a small device was connected to the customized PCB. A 200 mV peak-to-peak AC signal was applied to the device. The device was left on for more than 48 hours at ambient temperature. During the 48 hours, the DC voltage output was recorded for 8 times, and the average voltage output during each recording time was calculated, as shown in the third column of Table B.1.

		XZ C (1
File name	Time	V _{DC} of the
		PCB output
		(V)
device_nextday_PCB4_443`	10/23/2014	0.2083
+noGROUND_200m.lvm	4:43pm	
device_nextday_PCB4_444	10/23/2014	0.2079
+noGROUND_200m_10min.lvm	4:55pm	
device_nextday_PCB4_457	10/23/2014	0.2069
+noGROUND_200m_80min.lvm	6:18pm	
device_nextday_PCB4_834	10/23/2014	0.2031
+noGROUND_200m_2.5h.lvm	8:34pm	
device_nextday_PCB4_00.13	10/24/2014	0.2044
+noGROUND_200m_8h.lvm	00:13am	
device_nextday_PCB4_1024_1057	10/24/2014	0.2025
+noGROUND_3h.lvm	10:57am	
device_nextday_PCB4_1024_1128pm	10/24/2014	0.2031
+noGROUND_8h	11:28pm	
device_nextday_PCB4_1025_516pm	10/25/2014	0.1878
+noGROUND_10sec.lvm	5:16pm	

Table B.1 Voltage output from the PCB.

APPENDIX C

CALIBRATION OF ATI NANO17 LOAD CELL WITH WENGLOR® CP08MHT80 DISTANCE SENSOR

A cone-shaped structure with a weight of 136.4 g is assembled to the load cell. Without applying any dynamic load, the readouts from the load cell before and after the assembly are calculated as 0.00236 N and 1.3440 N, respectively. The difference in force equals to the weight of an object with a mass of 136.8 g. The difference is only 0.4 g, 0.29% of the nominal value, indicating the measurement of static force by ATI NANO17 load cell is very accurate.

The dynamic calibration of ATI NANO17 with CP08MHT80 is also conducted. The experimental setup is shown in Figure C.1, the cone-shaped structure is connected to the load cell which is directly connected to the vibration shaker. Table C.1 shows the experimental results when the peak to peak value of the power supply of the vibration shaker is equal to 1 V. The first column of Table C.1 shows the vibration frequency, the second column shows the vibration amplitude of the cone-shaped structure which represents the vibration amplitude provided by the shaker and is obtained from the distance sensor. The third column shows the amplitude calculated from the force sensor, and the forth column presents the theoretical value of the force amplitude, derived from Equations C 1-5.



Figure C.1 Setup for conducting dynamic calibration of the load cell.

Since the distance sensor is calibrated first, as elaborated in APPENDIX C we can assume that the motion captured by the Wenglor® CP08MHT80 distance sensor can truly represent the vibration from the shaker. Theoretically, the vibration motion from the vibration shaker can be represented as:

$$x = A\sin(\omega t) \tag{C.1}$$

The velocity and acceleration can be derived as:

$$\ddot{x} = -A\omega^2 \sin(\omega t) \tag{C.2}$$

$$\frac{F}{m} = \ddot{x} = -A\omega^2 \sin(\omega t) \tag{C.3}$$

Thus, the theoretical amplitude of the force can be written as

$$A_F = mA\omega^2 \tag{C.4}$$

The actual force signal can be represented as

$$F = A_2 \sin(\omega t + \phi) \tag{C.5}$$

where ϕ is the phase shift between the force signal and displacement signal, caused either by the sensor itself or the environment.

Frequency (Hz)	Amplitude of the distance	Amplitude of the force (N)	Theoretical value of force amplitude (N)	e
10	sensor (µm)	0.1024	0.1007	0.014
10	352.2	0.1924	0.1897	0.014
15	453.7	0.5595	0.5497	0.017
20	594.6	1.2363	1.2807	0.035
25	229.5	0.6911	0.7725	0.117
30	116.0	0.5913	0.5621	0.049
35	70.9	0.5205	0.4675	0.101
40	52.4	0.4802	0.4511	0.060
45	47.5	0.4707	0.5182	0.100
50	31.4	0.4573	0.4228	0.075
55	28.2	0.4409	0.4592	0.041

Table C.1 Comparison of force amplitudes between experimental and theoretical values under different vibration amplitude.

Ideally, A_F and A_2 should equal to each other, the last column provides the difference between the two in percentage, by applying Equation C.6.

$$e = \frac{|A_F - A_2|}{A_2} \times 100\%$$
 (C.6)

Overall, these two values show good agreement.

APPENDIX D

DERIVATION OF THE VOLTAGE OUTPUT EQUATION OF THE MICROFLUIDIC DEVICE

Figure D. 1 shows the configuration of the circuit that converts the resistance change of a sensing segment to a DC voltage.



Figure D.1 Configuration of the circuits.

The input of the microfluidic device is expressed as:

$$v_{ac} = \frac{v_{PP}}{2}\sin(\omega t) \tag{D.1}$$

where v_{PP} is the peak to peak value of the voltage output from the function generator.

After the first stage of the circuit: a transimpedance amplifier, the voltage becomes:

$$v'_{ac} = \frac{v_{PP}}{2} \cdot \frac{R_F}{R_i} \sin(\omega t)$$
 (D.2)

where R_F is the resistance of a feedback resistor, and R_i is the resistance of the *i*th sensing segment of the microfluidic device, which can be detected by the *i*th transducer.

Then the amplified voltage goes to the second stage of the circuit: a multiplier and the voltage after this stage equals to:

$$v''_{ac} = \frac{v_{PP}^2}{4} \cdot \frac{R_F^2}{R_i^2} \sin^2(\omega t)$$
 (D.3)

This expression can be further written into:

$$v'' = \frac{v_{PP}^2}{4} \cdot \frac{R_F^2}{R_i^2} \sin^2(\omega t) = \frac{v_{PP}^2}{8} \cdot \frac{R_F^2}{R_i^2} [1 - \cos(2\omega t)]$$
(D.4)

After the third stage of the circuit, a low pass filter stage, only the DC output remains.

$$V_{out} = \frac{v_{PP}^2}{8} \cdot \frac{R_F^2}{R_i^2}$$
 (D.5)

APPENDIX E

MEASUREMENT OF THE SPATIALLY-VATYING PHASE SHIFTS WITH A DEVICE WITH A LARGE TRANSDUCER SPACEING

The first measurement is carried out with the microfluidic device with a larger transducer spacing as shown, which is also referred to as device #1 in Chapter two. Figure E.1 shows the customized experimental setup. The device is mounted on a printed circuit board (PCB) which is further fixed on a 5-axis manipulator. A rigid cylinder probe with a dimension of 11mm in length and 0.8 mm in radius is connected to a load cell through a probe holder. The probe holder is connected to a vibration shaker (LDS V201, Brüel & Kjær Company) through a metal sleeve.



Figure E.1 Experimental setup for measuring the spatially-varying phase shifts.

A sample is placed in the liquid container for testing. For samples like agar, a phosphate buffered saline (PBS) solution is poured to a container built on the device for providing an aqueous condition. The frequency of the external load is controlled by a function generator (Hewlett Packard 33120A) connected to the shaker through a power amplifier. A piezoelectric type load cell (Kistler-9712B5) is assembled in the probe holder right above the probe to record the frequency and amplitude of the applied sinusoidal load.

Another function generator Agilent 33220A is used to provide a sinusoidal input signal for the microfluidic device. As mentioned in Chapter two, the device serves as an impedance component in the circuit. After the transimpedance amplifier and demodulation stages, the AC voltage input signal becomes DC voltage output. Thus, when the device is under sinusoidal mechanical stimuli, the output voltage signal exhibits the same frequency, and the strength is proportional to the mechanical stimuli.

The output signals include a voltage signal from the load cell and voltage signals from the five transducers of the device. All the signals feed into a DAQ board (NI PCI-6133, 14 bit, 2.5 MS/s/ch) and are recorded at a sampling rate of 5 kHz for low-frequency loads (<100Hz) and 10 kHz for high-frequency (>100Hz) loads, using a custom LabVIEW program.

In the first measurement, both homogeneous and heterogeneous PDMS and agar samples were prepared. The samples were prepared the same way as those were prepared for the CSM for the elastic modulus measurement in Chapter 4. For PDMS samples, 4 samples were prepared, two homogeneous samples and two heterogeneous samples. The dimension of the samples and the location of heterogeneity within each sample are listed in Table E.1 and Table E.2.

Sample	Dimensions(mm)
1:30 PDMS	10.16×6.35×4.46
1:40 PDMS	10×6×4.19
1% agar	10×5.5×4.14
2% agar	10×7×4.07
3% agar	10×6×3.29

Table E.1 Dimension of the homogeneous samples.

Table E.2 Dimension of the heterogeneous samples and the heterogeneity position.

Sample	Dimensions(mm)	Heterogeneity position
1:30 PDMS	10.16×6.35×3.80	In the middle
1:40 PDMS	10×6×3.81	Three particles, two big ones on the left
1% agar	9×8×5.01	In the middle
2% agar	10.5×7×5.99	In the middle
3% agar	10×6×3.08	In the middle but close to the left

With this setup, the frequency of the input signal varies from 5 Hz to 150 Hz. From the control experiment, not only the misalignment and microchannel variation are observed, the resonance of the device is also observed, so that the valid dynamic measurement of material samples is confined within a certain frequency range. From the data collection side, the sampling rate for frequencies lower than 100 Hz is set to be 5000, while at frequencies higher than 100 Hz, the sampling rate is set to be 10,000.

In the measurement, an initial 500 µm deformation is applied to each sample before a sinusoidal load is added. For homogeneous 1:30 PDMS samples, the frequency was added towards 170 Hz and for homogeneous 1:40 PDMS samples, the frequency was added up to 200 Hz. For the agar samples, 1% agar samples are tested below 110 Hz for both homogeneous and heterogeneous samples. 2% agar samples are tested below 140 Hz. For the 3% homogeneous sample, the frequency range is 1-140 Hz, and the 3% heterogeneous sample is tested every 10 Hz, up to 200 Hz.

Figure E.2 shows the dynamic characterization data of the device itself at the frequency of 10 Hz. As predicted, a phase lag exists between the load input and the voltage output signals from the transducers, due to the PDMS structure and the liquid within the microchannel. Only a slight phase shift is observed between the five voltage outputs, indicating the variations between the transducers are small. Figure E.3 shows the phase shift between the sinusoidal load input and the output voltage signals at different frequencies. As can be seen, the phase variation is relatively small at lower frequencies, and becomes larger at higher frequencies. The phase shift shows an increasing trend from the first to the fifth transducer, indicating the first transducer has a quicker response toward the external load, which further indicates the probe is tilted toward the first transducer.



Figure E.2 Control experiment data in time domain at 10 Hz (The dashed lines show the phase lag between the applied force and the voltage output).



Figure E.3 Measured phase shift between the load input and the voltage outputs of the device as a function of frequency.

From Figure E.3, it is observed that at around 80 Hz, an abrupt change takes place, due to the device resonance or because another mode different from the actuation mode kicks in. Thus, the dynamic measurement of the soft materials should be kept under 80 Hz for meaningful interpretation. This measurement is intended to serve as a preliminary study of the phase shift, further study on the system-level parameters are not calculated at this stage.

Figure E.4 and Figure E.5 plot the CSM results of 1:30 and 1:40 PDMS samples. It is very obvious that under 80 Hz, only very small phase variations exist among signals from different transducers, which indicates the homogeneous property of the test materials. Figure E.6, Figure E.7, and Figure E.8 show the dynamic measurement results of 1%, 2%, and 3% homogeneous agar samples, respectively. Due to the fragility of agar samples, limited data were collected for the 2% and 3% agar samples below the frequency of 80 Hz. For the 1% and 2% agar samples, the results

are very consistent. When the frequency is below 80 Hz, the variation of the phase shift among transducers are very small, which proves the homogeneity of the samples under test. For the 3% agar homogeneous sample, the phase variation among the first four transducers is very small, but the fifth transducer shows a smaller phase shift than the others. This possibly can be attributed to uneven thickness in some specific region.

Error bars are all added on the data points for each measurement. The maximum standard deviation of the measurement for different samples below 80 Hz are shown in Table E.3.

Table E.3 The maximum standard deviation of the measurement for different homogeneoussamples below 80 Hz.

Sample	1:30 PDMS	1:40 PDMS	1% agar	2% agar	3% agar
Maximum standard deviation (degree)	4.58	7.55	13.18	7.15	2.42



Figure E.4 Relation between phase shift and frequency of a 1:30 homogeneous PDMS sample.



Figure E.5 Relation between phase shift and frequency of a 1:40 homogeneous PDMS sample.



Figure E.6 Relation between phase shift and frequency of a 1% agar sample.



Figure E.7 Relation between phase shift and frequency of a 2% agar sample.



Figure E.8 Relation between phase shift and frequency of a 3% agar sample.

By comparing the curves in Figure E.9, it can be seen that the phase shifts of transducers 2, 3 and 4 of the 1:30 heterogeneous PDMS sample are much smaller than the rest, indicating the

viscous property of the material right above 2, 3 and 4 transducers are different from the rest. Figure E.10 shows the measurement result of a heterogeneous 1:40 PDMS sample. Transducers 1, 2 and 3 exhibit a smaller phase shift compared to transducers 4 and 5. There are three particles in the sample, two big ones are located on the left, making the neighboring area stiffer and less viscous than the rest. The result also indicates the portion of material above the first three transducers is less viscous than the rest, which is consistent with the composition of the sample.



Figure E.9 Relation between phase shift and frequency of a 1:30 heterogeneous PDMS sample.



Figure E.10 Relation between phase shift and frequency of a 1:40 heterogeneous PDMS sample.



Figure E.11 Relation between phase shift and frequency of a 1% heterogeneous agar sample.



Figure E.12 Relation between phase shift and frequency of a 2% heterogeneous agar sample.



Figure E.13 Relation between phase shift and frequency of a 3% heterogeneous agar sample.

Figure E.11 to Figure E.13 present the dynamic measurement results of 1%, 2%, and 3% heterogeneous agar samples. As can be seen from these phase-frequency plots, the results of the

agar samples are less stable than those of the PDMS samples. The agar samples were tested in a PBS solution, and are fragile compared with the PDMS samples, making it difficult to obtain stable results, especially for the 1% and 2% agar samples.

APPENDIX F

LABVIEW PROGRAMS FOR INSTRUMENT CONTROL AND DATA ACQUISITION



Figure F.1 LabVIEW block diagram for (a) controlling the micromanipulator to apply different levels of indentation depth to material samples and (b) the data collection.



(b)

Figure F.1. (Continued)



Figure F.2 LabVIEW block diagram for the shaker control to achieve constant vibration amplitude.

APPENDIX G

MATLAB CODES FOR DATA PROCESSING

```
% Nonlinear regression method to achieve the system-level parameters%
clc
clear all
close all
f=5:40;
z0=[9.91E-05 9.67E-05 0.00010712 0.00011034 0.00010782
0.000100114 9.62E-05 9.53E-05
                                   9.80E-05
                                                9.66E - 05
0.000103286 9.82E-05
                       9.62E-05
                                   9.51E-05
                                                9.73E-05
                                                            9.45E-05
                       9.33E-05
                                   8.86E-05
                                                8.93E-05
9.65E-05
            9.52E-05
                                                            9.32E-05
                       9.99E-05 0.000100897 9.30E-05
9.40E-05
            9.55E-05
                                                            9.55E-05
0.000100916 9.87E-05
                      9.91E-05 0.000100516 0.000100804 9.97E-05
9.92E-05
            9.90E-05
];
F0=[0.16836611 0.177534916 0.178110174 0.187912649 0.183247843
0.162031866 0.168414074 0.154902047 0.151326204 0.138722207
0.135624351 0.11585405 0.103642978 0.091467736 0.090283536
0.076590228 0.06727397 0.061779085 0.056405767 0.060778504
0.067281624 0.087091992 0.091436724 0.099011803 0.129618284
0.139042234 0.148400652 0.167522072 0.191482841 0.212552912
0.245278147 0.26773514 0.292510417 0.31893007 0.351961793
0.383674814
];
r=z0./F0;
coeff0=[200 0.1 9];
[coeff1,r1,J1,COVB1,mse1] = nlinfit(f',r',@fun,coeff0);
r2=1./sqrt((coeff1(1)-
coeff1(2)*(2*3.14159.*f).^2).^2+(coeff1(3)*2*3.14159.*f).^2);
plot(f,r,f,r2)
legend('experiement', 'fitting')
function ratio=fun(coeff,f)
k=coeff(1);
m = coeff(2);
D=coeff(3);
ratio=1./sqrt((k-m*(2*3.14159.*f).^2).^2+(D*2*3.14159.*f).^2);
```

close all % Codes for obtaining phase shift between force and voltage output % clc clear all names={ 'May 012015 656 PM try1 sine 10 0.965 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 656 PM try1 sine 11 0.91 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 657 PM try1 sine 12 0.88 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 658 PM try1 sine 13 0.827 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 658 PM try1 sine 14 0.761 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 659 PM try1 sine 15 0.729 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 659 PM try1 sine 16 0.629 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 704 PM try1 sine 17 0.627 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 709 PM try1 sine 18 0.484 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 712 PM try1 sine 19 0.455 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 716 PM try1 sine 20 0.421 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 723 PM try1 sine 21 0.344 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 723 PM try1 sine 22 0.621 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 728 PM try1 sine 23 0.604 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 730 PM try1 sine 24 0.644 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 733 PM try1 sine 25 0.655 5000SR DeviceEcoflex 200um_100kHz220mV.lvm',... 'May 012015 735 PM try1 sine 26 0.645 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 737 PM try1 sine 27 0.862 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 739 PM try1 sine 28 1.03 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 740 PM try1 sine 29 1.16 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 742 PM try1 sine 30 1.18 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 744 PM try1 sine 31 1.19 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',... 'May 012015 746 PM try1 sine 32 1.28 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...

```
'May 012015 748
PM try1 sine 33 1.40 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 750
PM try1 sine 34 1.55 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 751
PM try1 sine 35 1.70 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 753
PM try1 sine 36 1.86 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 756
PM try1 sine 37 2.04 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 800
PM try1 sine 38 2.21 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 802
PM try1 sine 39 2.43 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
    'May 012015 803
PM try1 sine 40 2.63 5000SR DeviceEcoflex 200um 100kHz220mV.lvm',...
       };
obj=10:40;
calibration matrix=[-0.007319839 0.009288162 0.162821761 -
1.694791278 -0.080131955 1.699079442
-0.174537551
                2.016933262 0.06230442 -0.983231488 0.03771666 -
0.980472141
1.920779685 0.032519337 1.920579462 0.035781451 1.933560047
0.015389205
               12.20066031 11.05287936 -5.731859987
                                                      -10.57384713
-1.159629186
-6.020251137
-12.06998174
                                5.193231312 10.39931369 6.930910878 -
               -0.296010123
10.24664593
-0.778311937
               7.062738378 -0.526085687 7.140892275 -0.15962376
7.141832434
];
LC initial=[-
0.0734515956399962,1.66005411407976,0.187463334879998,1.3197497673998
6,-0.0918308605600016,1.42308237107987];
for i=1:length(obj)
Data= dlmread(string(names[102]),',', 23, 0);
t=Data(:,1);
 SR=1/(t(2)-t(1));
 LC=Data(:,8:13);
 F 6=(calibration matrix*(LC-ones(size(LC))*diag(LC_initial))')';
 F z=F 6(:,3);
 disp=Data(:,7)/0.2*1e-3;
 for j=1:5
    x1(:,j)=Data(:,j+1);
    [Cxy(:,j),F(:,j)] = cpsd(F z,x1(:,j),hamming(1000),400,2^17,SR);
    angles (:, j) = -angle (Cxy(:, j));
    [c index(j)] = min(abs(F(:, j)-obj(i)));
    angleValues(j,i) =-angles(index(j),j)/pi*180;
```

```
drift(j,i)=mean(x1(:,j));
 end
 amp F sum(i)=0;
 amp disp sum(i)=0;
for j=1:obj(i)*5
amp_F_sum(i) = amp_F_sum(i)+((max(F_z(j*(i-1)+1:j*(i-
1)+floor(SR/obj(i))) ))- (min(F z(j*(i-1)+1:j*(i-1)+floor(SR/obj(i)))
)))/2;
amp_disp_sum(i) = amp_disp_sum(i)+((max(disp(j*(i-1)+1:j*(i-
1)+floor(SR/obj(i))) ))- (min(disp(j*(i-1)+1:j*(i-
1)+floor(SR/obj(i))))/2;
end
figure (i+1)
plot(obj,angleValues)
legend('1','2','3','4','5')
figure (i+2)
plot(obj,ratio)
```

Wenting Gu received her B.S. degree in Aerospace Engineering from Nanjing University of Aeronautics and Astronautics in 2009, and received her M.E. degree in Mechanical Engineering from Old Dominion University in 2013. She is currently a Ph.D. candidate in the Department of Mechanical and Aerospace Engineering at Old Dominion University. She has been working on the design, modeling, and characterization of MEMS/Micro sensors since 2010. Her research interests are developing MEMS/Micro sensors, solving multiphysics problems encountered by MEMS/Micro sensors, and exploring their applications.